## Features

- 80C51 Compatible
- Two I/O Ports
- Two 16-bit Timer/Counters
- 256 bytes RAM
- 4 Kbytes ROM or 4 Kbytes Flash Program Memory
- 256 bytes EEPROM (Stack Die Packaging Technology on SO20 Package)
- X2 Speed Improvement Capability ( 6 Clocks/Machine Cycle)
- 10-bit, 6 Channels A/D Converter
- One-channel with Progammable Gain and Rectifying Amplifier (Accuracy +/-5\%)
- Voltage Reference for A/D \& External Analog
- Hardware Watchdog Timer
- Programmable I/O Mode: Standard C51, Input Only, Push-pull, Open Drain
- Asynchronous Port Reset
- Triple System Clock
- Crystal or Ceramic Oscillator ( 24 MHz )
- RC Oscillator ( 12 MHz ), with Calibration Factor Using External R and C (Accuracy +/- $3.5 \%$ with Ideal R and C)
- RC Oscillator, Low Power Consumption (12 MHz Low Accuracy)
- Programmable Prescaler
- One PWM Unit Block With:
- 16-bits Programmable Counter
- 3 Independent Modules
- One PWM Unit Block with:
- 16 bits Programmable Counter
- 1 Module
- Interrupt Structure With:
- 7 Interrupt Sources,
- 4 interrupt Priority Levels
- Power Control Modes:
- Idle Mode
- Power-down Mode
- Power Fail Detect, Power On Reset
- Quiet mode for A to D Conversion
- Power Supply: 3 to 3.6 V
- Temperature Range: -40 to $85^{\circ} \mathrm{C}$
- Package: SO20, SO24 (upon request)


## Description

The AT8xEB5114 is a high performance version of the 80C51 8-bit microcontroller in a Low Pin Count package.
The AT8xEB5114 retains all the features of the standard 80C51 with 4 Kbytes program memory, 256 bytes of internal RAM, a 7 -source, 4 -level interrupt system, an onchip oscillator and two timers/counters. AT8xEB5114 may include a serial two wire interface EEPROM housed together with the microcontroller die in the same package.
The AT8xEB5114 is dedicated for analog interfacing applications. For this, it has a 10bit, 6 channels A/D converter and two PWM units; these PWM blocks provide PWM generation with variable frequency and pulse width.
In addition, the AT8xEB5114 has a Hardware Watchdog Timer and an X2 speed improvement mechanism. The X2 feature allows to keep the same CPU power at a divided by two oscillator frequency. The prescaler allows to decrease CPU and peripherals clock frequency. The fully static design of the AT8xEB5114 allows to reduce system power consumption by bringing the clock frequency down to any value, even DC, without loss of data.

The AT8xEB5114 has 3 software-selectable modes of reduced activity for further reduction in power consumption. In idle mode the CPU is frozen while the peripherals are still operating. In quiet mode, only the A/D converter is operating. In power-down mode the RAM is saved and all other functions are inoperative. Three oscillator sources, crystal, precision RC and low power RC, provide versatile power management.

The AT8xEB5114 is available in low pin count packages (ROM and flash versions).
Figure 1. Block Diagram


## Pin Configuration

| P4.0/AINO/W0CI 1 |  | 20 | 7 VRef |
| :---: | :---: | :---: | :---: |
| P4.1/AIN1/T1 2 |  | 19 | Vcca |
| P4.2/AIN2/W1CI 3 |  | 18 | $\square \mathrm{Vssa}$ |
| P4.3/AIN3/INT1 4 |  | 17 | R |
| P3.3/W0M2/AIN4 [5 |  | 16 | C |
| P3.4/T0/AIN5 6 | SO20 | 15 | XTAL2 |
| P3.5/W1M0 7 |  | 14 | XTAL1 |
| P3.2/INT0 0 |  | 13 | $\overline{\mathrm{RST}}$ |
| P3.1/W0M1 9 |  | 12 | Vss |
| P3.0/W0M0 10 |  | 11 | ] Vcc |


| P4.0/AIN0/W0CI 1 |  | 24 | VRef |
| :---: | :---: | :---: | :---: |
| P4.1/AIN1/T1 2 |  | 23 | Vcca |
| P4.2/AIN2/W1CI 3 |  | 22 | Vssa |
| P4.3/AIN3/INT1 4 |  | 21 | NC |
| P3.3/W0M2/AIN4 [5 |  | 20 | R |
| P3.4/T0/AIN5 6 | SO24 | 19 | C |
| P3.6 7 | No EE | 18 | XTAL2 |
| P3.5/W1M0 8 |  | 17 | XTAL1 |
| P3.2/INT0 9 |  | 16 | NC |
| P3.1/W0M1 10 |  | 15 | $\overline{\mathrm{RST}}$ |
| P3.0/W0M0 11 |  | 14 | Vss |
| P3.7 12 |  | 13 | Vcc |

Pin Description

| SO20 | SO24 | Mnemonic | Type | Name and Function |
| :---: | :---: | :---: | :---: | :---: |
| 12 | 14 | $\mathrm{V}_{\text {S }}$ | Power | Ground: OV reference |
| 18 | 22 | Vssa | Power | Analog Ground: 0V reference for analog part |
| 11 | 13 | $\mathrm{V}_{\text {cc }}$ | Power | Power Supply: This is the power supply voltage for normal, idle and power-down operation. |
| 19 | 23 | Vcca | Power | Analog Power Supply: This is the power supply voltage for analog part This pin must be connected to power supply. |
| 20 | 24 | VREF | Analog | VREF: A/D converter positive reference input, output of the internal voltage reference |
| 14 | 17 | XTAL1 | 1 | Input to the inverting oscillator amplifier and input to the internal clock generator circuit |
| 15 | 18 | XTAL2 | 0 | Output from the inverting oscillator amplifier. This pin can't be connected to the ground. |
| 17 | 20 | R | Analog | Resistor Input for the precision RC oscillator |
| 16 | 19 | C | Analog | Capacitor Input for the precision RC oscillator |
| 13 | 15 | $\overline{\mathrm{RST}}$ | I/O | Reset input with integrated pull-up <br> A low level on this pin for two machine cycles while the oscillator is running, resets the device. |
|  |  | P3.0-P3.7 | I/O | Port 3: Port 3 is an 8 -bit programmable I/O port with internal pull-ups. See "Port Types" on page 32. for a description of I/O ports. <br> Port 3 also serves the special features of the 80C51 family, as listed below. |
| 10 | 11 |  | I/O | W0M0 (P3.0): External I/O for PWMU 0 module 0 |
| 9 | 10 |  | 1/O | W0M1 (P3.1): External I/O for PWMU 0 module 1 |
| 8 | 9 |  | I/O | $\overline{\text { INTO }}$ (P3.2): External interrupt 0 |
| 5 | 5 |  | I/O | WOM2 / AIN4 (P3.3): External I/O for PWMU 0 module 2. P3.3 is also an input of the analog to digital converter. |
| 6 | 6 |  | 1/O | T0 / AIN5(P3.4): Timer 0 external input. P3.4 is also an input of the analog to digital converter. |
| 7 | 8 |  | I/O | W1M0 (P3.5): External I/O for PWMU 1 module 0, can also be used to output the external clocking signal |
|  |  | P4.0-P4.3 | I/O | Port 4: Port 4 is an 4-bit programmable I/O port with internal pull-ups. See "Port Types" on page 32. for a description of I/O ports. <br> Port 4 is also the input port of the Analog to digital converter |
| 1 | 1 |  | I/O | AINO (P4.0): A/D converter input 0 <br> W0CI: Count input of PWMU0 |
| 2 | 2 |  | I/O | AIN1 (P4.1): A/D converter input 1 <br> T1: Timer 1 external input |
| 3 | 3 |  | I/O | AIN2 (P4.2): A/D converter input 2 <br> W1CI: Count input of PWMU1 |
| 4 | 4 |  | I/O | AIN3 (P4.3): A/D converter input 3, programmable gain INT1: External interrupt 1 |

## SFR Mapping

The Special Function Registers (SFRs) of the AT8xEB5114 belong to the following categories:

- C51 core registers: ACC, AUXR, AUXR1, B, DPH, DPL, PSW, SP, FCON, HSB
- I/O port registers: P3, P4, P3M1, P3M2, P4M1
- Timer registers: TCON, TH0, TH1, TL0, TL1, TMOD
- Power and clock control registers: CKCON, CKRL, CKSEL, OSCBFA, OSCCON, PCON
- Interrupt system registers: IENO, IPHO, IPLO, IOR
- WatchDog Timer: WDTRST, WDTPRG
- PWMO registers: WOCH, WOCL, WOCON, WOFH, WOFL, WOIC, WOMOD, WOROH, W0R0L, WOR1H, W0R1L,W0R2H, W0R2L
- PWM1registers: W1CH, W1CL, W1CON, W1FH, W1FL, W1IC, W1R0H, W1R0L
- ADC registers: ADCA, ADCF, ADCLK, ADCON, ADDH, ADDL

Table 1. SFR Addresses and Reset Values

| F8h | 0/8 | 1/9 | 2/A | 3/B | 4/C | 5/D | 6/E | 7/F | FFh |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & \text { W1CON } \\ & \text { XXX0 } 0000 \end{aligned}$ |  | W1FH 00000000 | W1FL 00000000 | $\begin{gathered} \text { W1CH } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { W1CL } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { W1IC } \\ 00000000 \end{gathered}$ |  |  |
| F0h | $\begin{gathered} \text { B } \\ 00000000 \end{gathered}$ |  | $\begin{gathered} \text { ADCLK } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { ADCON } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { ADDL } \\ \text { XXXXX00 } \end{gathered}$ | $\begin{gathered} \text { ADDH } \\ 00000000 \end{gathered}$ | $\begin{aligned} & \text { ADCF } \\ & 00000000 \end{aligned}$ | $\begin{gathered} \text { ADCA } \\ 00000000 \end{gathered}$ | F7h |
| E8h | $\begin{aligned} & \text { WOCON } \\ & \text { 00XX } 0000 \end{aligned}$ | $\begin{aligned} & \text { WOMOD } \\ & \text { 00XX X000 } \end{aligned}$ | WOFH 00000000 | WOFL 00000000 | $\begin{gathered} \text { WOCH } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { WOCL } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { WOIC } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { HSB } \\ 1111 \text { XX11 } \end{gathered}$ | EFh |
| E0h | $\begin{gathered} \text { ACC } \\ 00000000 \end{gathered}$ |  |  |  | $\begin{gathered} \text { P3M2 } \\ 00000000 \end{gathered}$ |  |  |  | E7h |
| D8h |  | $\begin{gathered} \text { WOROH } \\ 00000000 \end{gathered}$ | WOROL 00000000 | $\begin{aligned} & \text { W0R1H } \\ & 00000000 \end{aligned}$ | $\begin{aligned} & \text { W0R1L } \\ & 00000000 \end{aligned}$ | $\begin{aligned} & \text { WOR2H } \\ & 00000000 \end{aligned}$ | $\begin{gathered} \text { W0R2L } \\ 00000000 \end{gathered}$ |  | $\begin{gathered} \text { DF } \\ \mathrm{h} \end{gathered}$ |
| DOh | $\begin{gathered} \text { PSW } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { FCON } \\ 11111111 \end{gathered}$ |  |  |  | $\begin{gathered} \text { P3M1 } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { P4M1 } \\ 00000000 \end{gathered}$ |  | D7h |
| C8h |  | $\begin{aligned} & \text { W1R0H } \\ & 00000000 \end{aligned}$ | W1ROL 00000000 |  |  |  |  |  | $\begin{gathered} \mathrm{CF} \\ \mathrm{~h} \end{gathered}$ |
| COh | $\begin{gathered} \text { P4 } \\ \text { XXXX } 1111 \end{gathered}$ |  |  |  |  |  |  |  | C7h |
| B8h | $\begin{gathered} \text { IPLO } \\ \times 0000000 \end{gathered}$ |  |  |  |  |  |  |  | BFh |
| B0h | $\begin{gathered} \text { P3 } \\ 11111111 \end{gathered}$ |  |  |  |  |  |  | $\begin{gathered} \text { IPHO } \\ \times 0000000 \end{gathered}$ | B7h |
| A8h | $\begin{gathered} \text { IENO } \\ 00000000 \end{gathered}$ |  |  |  |  |  |  |  | AFh |
| AOh |  |  | $\begin{gathered} \text { AUXR1 } \\ \text { XXXX 0XX0 } \end{gathered}$ |  |  | IOR <br> XXXXXX00 | WDTRST XXXXXXXX | WDTPRG XXXX X000 | A7h |
| 98h |  |  |  |  |  |  |  | $\begin{aligned} & \text { OSCBFA } \\ & 01110110 \end{aligned}$ | 9Fh |
| 90h |  |  |  |  |  |  |  | $\begin{gathered} \text { CKRL } \\ \text { XXXX } 1000 \end{gathered}$ | 97h |
| 88h | $\begin{gathered} \text { TCON } \\ 00000000 \end{gathered}$ | TMOD 00000000 | $\begin{gathered} \text { TLO } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { TL1 } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { TH0 } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { TH1 } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { AUXR } \\ \text { OXXO XXXO } \end{gathered}$ | $\begin{gathered} \text { CKCON } \\ \text { XXXX XXX0 } \end{gathered}$ | 8Fh |
| 80h |  | $\begin{gathered} \text { SP } \\ 00000111 \end{gathered}$ | $\begin{gathered} \text { DPL } \\ 00000000 \end{gathered}$ | $\begin{gathered} \text { DPH } \\ 00000000 \end{gathered}$ |  | $\begin{gathered} \text { CKSEL } \\ \text { XXXX XXCC } \end{gathered}$ | $\begin{gathered} \text { OSCCON } \\ \text { XXXX XXCC } \end{gathered}$ | $\begin{gathered} \text { PCON } \\ 00 \mathrm{XX} \text { XX00 } \end{gathered}$ | 87h |
|  | 0/8 | 1/9 | 2/A | 3/B | 4/C | 5/D | 6/E | 7/F |  |

Note: 1. "C", value defined by the Hardware Security Byte, see Table 2 on page 15

| Mnemonic | Add | Name | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ACC | E0h | Accumulator |  |  |  |  |  |  |  |  |
| ADCA | F7h | ADC Amplifier Configuration | - | - | - | - | - | AC3E | AC3G1 | AC3G0 |
| ADCF | F6h | ADCF Register | - | - | CH5 | CH 4 | CH3 | CH2 | CH1 | CH0 |
| ADCLK | F2h | ADC Clock Prescaler | SELREF | PRS6 | PRS5 | PRS4 | PRS3 | PRS2 | PRS1 | PRSO |
| ADCON | F3h | ADC Control Register | QUIETM | PSIDLE | ADEN | ADEOC | ADSST | SCH2 | SCH1 | SCHO |
| ADDH | F5h | ADC Data High Byte Register | ADAT9 | ADAT8 | ADAT7 | ADAT6 | ADAT5 | ADAT4 | ADAT3 | ADAT2 |
| ADDL | F4h | ADC Data Low Byte Register | - | - | - | - | - | - | ADAT1 | ADAT0 |
| AUXR | 8Eh | Auxiliary Register | DPU | - | - | LOWVD | - | - | - | - |
| AUXR1 | A2h | Auxiliary Register 1 | - | - | - | - | - | - | - | DPS |
| B | F0h | B Register |  |  |  |  |  |  |  |  |
| CKCON | 8Fh | Clock control Register | - | - | - | - | - | - | - | X2 |
| CKRL | 97h | Clock Prescaler Register | - | - | - | - | CKRL3 | CKRL2 | CKRL1 | CKRLO |
| CKSEL | 85h | Clock Selection register | - | - | - | - | - | - | CKS1 | CKSO |
| DPH | 83h | Data pointer High Byte |  |  |  |  |  |  |  |  |
| DPL | 82h | Data pointer Low Byte |  |  |  |  |  |  |  |  |
| FCON | D1h | Auxiliary Register | FPL3 | FPL2 | FPL1 | FPLO | FPS | FMOD1 | FMOD0 | FBUSY |
| HSB | EFh | Hardware Security Byte | X2 | RST_OSC1 | RST_OSC0 | RST_OCLK | - | - | LB1 | LB0 |
| IENO | A8h | Interrupt Enable Register | EA | EADC | EW1 | EW0 | ET1 | EX1 | ETO | EX0 |
| IOR | A5h | Interrupt Option Register | - | - | - | - | - | - | ESB1 | ESB0 |
| IPH0 | B7h | Interrupt Priority register | - | PADCH | PW1H | PWOH | PT1H | PX1H | PTOH | PXOH |
| IPLO | B8h | Interrupt Priority Register | - | PADC | PW1 | PW0 | PT1 | PX1 | PT0 | PX0 |
| OSCBFA | 9Fh | Oscillator B Frequency Adjust | OSCBFA7 | OSCBFA6 | OSCBFA5 | OSCBFA4 | OSCBFA3 | OSCBFA2 | OSCBFA1 | OSCBFAO |
| OSCCON | 86h | Clock Control Register | - | - | - | OSCBRY | LCKEN | OSCCEN | OSCBEN | OSCAEN |
| P3 | B0h | Port 3 Register |  |  |  |  |  |  |  |  |
| P3M1 | D5h | Port 3 Output Configuration | P3M1.7 | P3M1.6 | P3M1.5 | P3M1.4 | P3M1.3 | P3M1. 2 | P3M1.1 | P3M1.0 |
| P3M2 | E4h | Port 3 Output Configuration | P3M2.7 | P3M2.6 | P3M2.5 | P3M2.4 | P3M2.3 | P3M2.2 | P3M2.1 | P3M2.0 |
| P4 | COh | Port 4 register |  |  |  |  |  |  |  |  |
| P4M1 | D6h | Port 4 Output Configuration | P4M1.7 | P4M1.6 | P4M1.5 | P4M1.4 | P4M1.3 | P4M1.2 | P4M1. 1 | P4M1.0 |
| PCON | 87h | Power Modes Control Register | SMOD1 | SMOD0 | - | - | GF1 | GF0 | PD | IDL |
| PSW | DOh | Program Status Word | CY | AC | F0 | RS1 | RS0 | OV | F1 | P |
| SP | 81h | Stack pointer |  |  |  |  |  |  |  |  |
| TCON | 88h | Timer/Counter Control Register | TF1 | TR1 | TF0 | TR0 | IE1 | IT1 | IE0 | IT0 |
| TH0 | 8Ch | Timer 0 High Byte Registers | TH0.7 | TH0.6 | TH0.5 | TH0.4 | TH0.3 | TH0.2 | TH0.1 | TH0.0 |
| TH1 | 8Dh | Timer 1 High Byte Registers | TH1.7 | TH1.6 | TH1.5 | TH1.4 | TH1.3 | TH1.2 | TH1. 1 | TH1.0 |
| TLO | 8Ah | Timer 0 Low Byte Registers | TL0.7 | TL0.6 | TL0.5 | TLO. 4 | TL0.3 | TL0.2 | TL0.1 | TL0.0 |
| TL1 | 8Bh | Timer 1 Low Byte Registers | TL1.7 | TL1.6 | TL1.5 | TL1.4 | TL1.3 | TL1.2 | TL1. 1 | TL1.0 |


| Mnemonic | Add | Name | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TMOD | 89h | Timer/Counter Mode Register | GATE1 | C/T1\# | M11 | M01 | GATE0 | C/T0\# | M10 | M00 |
| WOCH | ECh | PWMUO Counter High Control | W0C15 | W0C14 | W0C13 | W0C12 | W0C11 | W0C10 | W0C9 | W0C8 |
| WOCL | EDh | PWMU0 Counter Low Control | W0C7 | W0C6 | W0C5 | W0C4 | W0C3 | W0C2 | W0C1 | W0C0 |
| WOCON | E8h | PWMU0 Control Register | WOUP | WOR | - | - | woos | W0EN2 | W0EN1 | WOENO |
| W0FH | EAh | PWMUO Frequency High Control | W0F15 | W0F14 | W0F13 | W0F12 | W0F11 | W0F10 | W0F9 | W0F8 |
| W0FL | EBh | PWMU0 Frequency Low Control | W0F7 | W0F6 | W0F5 | W0F4 | W0F3 | W0F2 | W0F1 | W0F0 |
| WOIC | EEh | PWMUO Interrupt Configuration | WOCF | W0CF2 | W0CF2 | WOCFO | WOECF | W0ECF2 | W0ECF1 | W0ECF0 |
| WOMOD | E9h | PWMUO Counter Mode Register | W0CPS1 | W0CPS0 | - | - | - | W0INV2 | W0INV1 | W0INV0 |
| WOROH | D9h | PWMU0 Module 0 High Toggle | W0ROH15 | W0ROH14 | W0ROH13 | W0ROH12 | W0R0H11 | W0ROH10 | W0ROH9 | WOROH8 |
| WOROL | DAh | PWMU0 Module 0 Low Toggle | W0ROH7 | W0ROH6 | W0ROH5 | W0ROH4 | W0ROH3 | WOROH2 | W0R0H1 | WOROHO |
| W0R1H | DBh | PWMU0 Module 1High Toggle | W0R1H15 | W0R1H14 | W0R1H13 | W0R1H12 | W0R1H11 | W0R1H10 | W0R1H9 | W0R1H8 |
| W0R1L | DCh | PWMU0 Module1 Low Toggle | W0R1H7 | W0R1H6 | W0R1H5 | W0R1H4 | W0R1H3 | W0R1H2 | W0R1H1 | W0R1H0 |
| W0R2H | DDh | PWMU0 Module 2 High Toggle | W0R2H15 | W0R2H14 | W0R2H13 | W0R2H12 | W0R2H11 | W0R2H10 | W0R2H9 | W0R2H8 |
| W0R2L | DEh | PWMU0 Module 2 Low Toggle | W0R2H7 | W0R2H6 | W0R2H5 | W0R2H4 | W0R2H3 | W0R2H2 | W0R2H1 | W0R2H0 |
| W1CH | FCh | PWMU1 Counter High Control | W1C15 | W1C14 | W1C13 | W1C12 | W1C11 | W1C10 | W1C9 | W1C8 |
| W1CL | FDh | PWMU1 Counter Low Control | W1C7 | W1C6 | W1C5 | W1C4 | W1C3 | W1C2 | W1C1 | W1C0 |
| W1CON | F8h | PWMU1 Control Register | W1UP | W1R | - | W1OCLK | W1CPS1 | W1CPS0 | W1INV0 | W1EN0 |
| W1FH | FAh | PWMU1 Frequency High Control | W1F15 | W1F14 | W1F13 | W1F12 | W1F11 | W1F10 | W1F9 | W1F8 |
| W1FL | FBh | PWMU1 Frequency Low Control | W1F7 | W1F6 | W1F5 | W1F4 | W1F3 | W1F2 | W1F1 | W1F0 |
| W1IC | FEh | PWMU1 Interrupt Configuration | W1CF | - | - | W1CF0 | W1ECOF | - | - | W0ECFO |
| W1R0H | C9h | PWMU1 Module 0 High Toggle | W1R0H15 | W1R0H14 | W1R0H13 | W1R0H12 | W1R0H11 | W1R0H10 | W1R0H9 | W1R0H8 |
| W1R0L | CAh | PWMU1 Module 0 Low Toggle | W1R0H7 | W1R0H6 | W1R0H5 | W1R0H4 | W1R0H3 | W1R0H2 | W1R0H1 | W1R0H0 |
| WDTRST | A6h | Watchdog Timer enable Register |  |  |  |  |  |  |  |  |
| WDTPRG | A7h | WatchDog Timer Duration Prg | - | - | - | - | - | S2 | S1 | S0 |

## Power Monitor

## Description

The Power Monitor function supervises the evolution of the voltages feeding the microcontroller, and if needed, suspends its activity when the detected value is out of specification.
It warrants proper startup when AT8xEB5114 is powered up and prevents code execution errors when the power supply becomes lower than the functional threshold.

This chapter describes the functions of the power monitor.

In order to startup and to properly maintain the microcontroller operation, Vcc has to be stabilized in the Vcc operating range and the oscillator has to be stabilized with a nominal amplitude compatible with logic threshold.
In order to be sure the oscillator is stabilized, there is an internal counter which maintains the reset during 1024 clock periods in case the oscillator selected is the OSC A and 64 clock periods in case the oscillator used is OSC B or OSC C.

This control is carried out during three phases: the power-up, normal operation and stop. In accordance with the following requirements:

- it guarantees an operational Reset when the microcontroller is powered-up, and
- a protection if the power supply goes below minimum operating Vcc

Figure 2. Power Monitor Block Diagram


## Power Monitor diagram

The Power Monitor monitors the power-supply in order to detect any voltage drops which are not in the target specification. The power monitor block verifies two kinds of situation that may occur:

- during the power-up condition, when Vcc reaches the product specification,
- during a steady-state condition, when Vcc is at nominal value but disturbed by any undesired voltage drops.

Figure 2 shows some configurations which can be handled by the Power Monitor.

Figure 3. Power-Up and Steady-state Conditions Monitored


The POR/PFD forces the CPU into reset mode when VCC reaches a voltage condition which is out of specification.
The thresholds and their functions are:

- VPFDP: the Vcc has reached a minimum functional value at power-up. The circuit leaves the RESET mode
- VPFDM: the Vcc has reached a low threshold functional value for the microcontroller. An internal RESET is set.

Glitch filtering prevents the system from RESET when short duration glitches are carried on Vcc power-supply (See "Electrical Characteristics" on page 84.).

In case Vcc is below VPFDP, LOWVD bit in AUXR (See Table 12 on page 23) is cleared by hardware. This bit allows the user to know if the voltage is below VPFDP.

Note: For proper reset operation $\mathrm{V}_{\mathrm{CCA}}$ and $\mathrm{V}_{\mathrm{CC}}$ must be considered together (same power source). However, to improve the noise immunity, it is better to have two decoupling networks close to power pins (one for $\mathrm{V}_{\mathrm{CCA}} / \mathrm{V}_{\mathrm{SSA}}$ pair and one for $\mathrm{V}_{\mathrm{CC}} / \mathrm{V}_{\mathrm{SS}}$ pair).

## Clock System

## Overview

## Blocks Description

The AT8xEB5114 oscillator system provides a reliable clocking system with full mastering of speed versus CPU power trade-off. Several clock sources are possible:

- External clock input
- High speed crystal or ceramic oscillator
- Integrated accurate oscillator with external R and C.
- Low power consumption Integrated RC oscillator without external components.

The AT8xEB5114 needs 6 clock periods per machine cycle when the X2 function is set. However, the selected clock source can be divided by 2-32 before clocking the CPU and the peripherals.

By default, the active oscillator after reset is the high speed crystal/ceramic oscillator. Any two bits in a hardware configuration byte programmed by a Flash programmer or by metal mask can activate any other one.
The clock system is controlled by several SFR registers: CKCON, CKSEL, CKRL, OSCON, PCON and HSB which is the hardware security byte.

The AT8xEB5114 includes three oscillators:

- Crystal oscillator optimized for 24 MHz .
- 1 accurate oscillator with a typical frequency of 12 MHz .
- 1 low power oscillator with a typical frequency of 14 MHz .

Figure 4. Functional Block Diagram


Crystal Oscillator: OSCA

High Accurate RC Oscillator: OSCB

The crystal oscillator uses two external pins, XTAL1 for input and XTAL2 for output. OSCAEN in OSCCON register is an enable signal for the crystal oscillator or for the external oscillator input that can be provided on XTAL1.

The high accuracy RC oscillator needs external R and C components to assure the proper accuracy; its typical frequency is 12 MHz . Frequency accuracy is a function of external $R$ and $C$ accuracy. It is recommended to use $0.5 \%$ or better for $R$ and $1 \%$ for C components. (Typical values are $\mathrm{R}=49.9 \mathrm{~K}$ and $\mathrm{C}=560 \mathrm{pF}$ )
This oscillator has two modes.

- OSCBEN = 1 and LCKEN = 0: Standard accuracy mode(Typical frequency 12 MHz )
- $\quad$ OSCBEN = 1 and LCKEN = 1: High accuracy mode (Typical frequency 12 MHz ). The OSCB oscillator is based on a low frequency RC oscillator and a VCO. When locked, the oscillator frequency is defined by the following formula:
$\mathrm{F}=3^{*}[\mathrm{OSCBFA}+1] /($ R.C). with C including parasitic capacitances.
Because the oscillator is based on a PLL, it needs several periods to reach its final accuracy. As soon as this accuracy is reached, the OSCBRY bit in OSCCON register is set by hardware.
The internal frequency is locked on the external RC time constant. So it is possible to adjust frequency by lower than $1 \%$ steps with the OSCBFA register. However the frequency adjustment is limited to +/-15\% around 12 MHz .
The frequency can be adjusted until $15 \%$ around 12 MHz by OSCBFA Register.

OSCBEN and LCKEN are in the OSCCON register.
The low power consumption RC oscillator doesn't need any external components. Moreover its consumption is very low. Its typical frequency is 14 MHz . Note that this on-chip oscillator has a $+/-40 \%$ frequency tolerance and may not be suitable for use in certain applications.
OSCC is set by OSCCEN bit in OSCCON.
CKS1 and CKS0 bits in CKSEL register are used to select the clock source.
OSCCEN bit in OSCCON register is used to enable the low power consumption RC oscillator.

OSCBEN bit in OSCCON register is used to enable the high accurate RC oscillator.
OSCAEN bit in OSCCON register is used to enable the crystal oscillator or the external oscillator input.

The AT8xEB5114 core needs only 6 clock periods per machine cycle. This feature called "X2" provides the following advantages:

- Divides frequency crystals by 2 (cheaper crystals) while keeping same CPU power.
- Saves power consumption while keeping same CPU power (oscillator power saving).
- Saves power consumption by dividing dynamically the operating frequency by 2 in operating and idle modes.
- Increases CPU power by 2 while keeping same crystal frequency.

In order to keep the original C51 compatibility, a divider by 2 is inserted between the XTAL1 signal and the main clock input of the core (phase generator). This divider may be enabled or disabled by software.

## Description

The clock for the whole circuit and peripherals is first divided by two before being used by the CPU core and the peripherals.

This allows any cyclic ratio to be accepted on XTAL1 input. In X2 mode, as this divider is bypassed, the signals on XTAL1 must have a cyclic ratio from 40 to $60 \%$.

Figure 4 shows the clock generation block diagram. X2 bit is validated on the rising edge of the XTAL1 $\div 2$ to avoid glitches when switching from X2 to standard mode. Figure 5 shows the switching mode waveforms.

Figure 5. Mode Switching Waveforms


The X2 bit in the CKCON register (see Table 7 on page 18) allows to switch from 12 clock periods per instruction to 6 clock periods and vice versa.

Before supplying the CPU and the peripherals, the main clock is divided by a factor from 2 to 32 , as defined by the CKRL register (see Table 6 on page 18). The CPU needs from 12 to $16^{*} 12$ clock periods per instruction. This allows:

- to accept any cyclic ratio on XTAL1 input.
- to reduce CPU power consumption.

Note: The number of bits of the prescaler is optimized in order to provide a low power consumption in low speed mode (see Section "Electrical Characteristics", page 84).

A hardware RESET selects the start oscillator depending on the RST1_OSC and RSTO_OSC bits contained on the Hardware Security Byte register (see Table 2 on page 15). It also selects the prescaler divider as follows:

- CKRL = 8h: internal clock = OscOut / 16 (slow CPU speed at reset, thus lower power consumption)
- $\mathrm{X} 2=0$,
- SEL_OSC1 and SEL_OSC0 bits selects OSCA, OSCB or OSCC, depending on the value of the RST_OSC1 and RST_OSC0 configuration bits.
- After Reset, any value between Fh down to Oh can be written by software into CKRL sfr in order to divide frequency of the selected oscillator:
- CKRL = Oh: minimum frequency = OscOut / 32
- CKRL = Fh: maximum frequency = OscOut / 2

The frequency of the CPU and peripherals clock CkOut is related to the frequency of the main oscillator OscOut by the following formula:
$F_{\text {CKOut }}=F_{\text {OscOut }} /\left(32-2^{*}\right.$ CKRL)

Some examples can be found in the table below:

| $\mathbf{F}_{\text {Oscout }}$ <br> $\mathbf{M H z}$ | $\mathbf{X 2}$ | CKRL | $\mathbf{F}_{\text {ckOut }}$ <br> $\mathbf{M h z}$ |
| :---: | :---: | :---: | :---: |
| 12 | 0 | F | 6 |
| 12 | 0 | E | 3 |
| 12 | 1 | x | 12 |

- A software instruction which set X2 bit disables the prescaler/divider, so the internal clock is either OSCA, OSCB or OSCC depending on SEL_OSC1 and SEL_OSC0 bits.


## Registers

Hardware Security Byte
The security byte sets the starting microcontroller options and the security levels.
The default options are X1 mode, Oscillator A and divided by 16 prescaler.
Table 2. Hardware Security Byte (HSB)
Power configuration Register - HSB (S:EFh)

| 7 | 6 |  | 5 |  | $\mathbf{4}$ | $\mathbf{3}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{7}$ | $\mathbf{2}$ | 1 | 0 |  |  |  |
| X2 | RST_OSC1 | RST_OSC0 | RST_OCLK | CKRLRV | - | LB1 |
| LB0 |  |  |  |  |  |  |


| $\begin{gathered} \text { Bit } \\ \text { Number } \end{gathered}$ | Bit Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | X2 | X2 Mode <br> Clear to force X2 mode (CkOut = OscOut) <br> Set to use the prescaler mode (CkOut = OscOut / (2*(16-M))) |
| 6 | RST_OSC1 | Oscillator bit 1 on reset and Oscillator bit 0 on reset <br> 11: allows OSCA <br> 10: allows OSCB <br> 01: allows OSCC <br> 00: reserved |
| 5 | RST_OSC0 |  |
| 4 | RST_OCLK | Output clocking signal after RESET <br> Clear to start the microcontroller with a low level on P3.5 followed by an output clocking signal on P3.5 as soon as the microcontroller is started. This signal has is a $1 / 3$ high $2 / 3$ low signal. Its frequency is equal to (CKout / 3). <br> Set to start on normal conditions: No signal on P3.5 which is pulled up. |
| 3 | CKRLRV | CKRL Reset Value <br> If set, the microcontroller starts with the prescaler reset value $=$ XXXX 1000 (OscOut = CkOut/16). <br> If clear, the microcontroller starts with a prescaler reset value $=$ XXXX 1111 (OscOut = CkOut/2). |
| 2 | - | Reserved |
| 1-0 | LB1-0 | User Program Lock Bits See Table 61 on page 81 |

HSB = 1111 1X11b

Clock Control Register
The clock control register is used to define the clock system behavior.
Table 3. OSCON Register
OSCCON - Clock Control Register (86h)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | OSCARY | OSCBRY | LCKEN | OSCCEN | OSCBEN | OSCAEN |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :--- |
| 7 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 6 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 5 | OSCARY | Oscillator A Ready <br> When set, this bit indicates that Oscillator A is ready to be used. |
| 4 | OSCBRY | Oscillator B Ready <br> When set, this bit indicates that Oscillator B is ready to be used in high accurate <br> mode. |
| 3 | LCKEN | Lock Enable <br> When set, this bit allows to increase the accuracy of OSCB by locking this <br> oscillator on external RC time constant. |
| 2 | OSCCEN | Enable low power consumption RC oscillator <br> This bit is used to enable the low power consumption oscillator <br> $0:$ The oscillator is disabled <br> $1:$ The oscillator is enabled. |
| 1 | OSCBEN | Enable high accuracy RC oscillator <br> This bit is used to enable the high accurate RC oscillator <br> $0:$ The oscillator is disabled <br> $1:$ The oscillator is enabled. |
| 0 | OSCAEN | Enable crystal oscillator <br> This bit is used to enable the crystal oscillator <br> $0:$ The oscillator is disabled <br> $1:$ The oscillator is enabled. |
| 2 |  |  |

Reset Value = XXX0
0"RST_OSC1.RST_OSC0""RST_OSC1. $\overline{R S T}$ _OSC0""RST_OSC1.RST_OSC0" b
Not bit addressable
Note: Before changing oscillator selection in CKSEL, be sure that the oscillator you select is started. OSCA is ready as soon as OSCARY is set by hardware, OSCB and OSCC are ready after 4 clock periods. In case you want to use OSCB locked, be sure that OSCB is started before setting LCKEN bit. Then, wait until OSCBRY is set by hardware to be sure that the accurate frequency is reached.

Oscillator B Frequency Adjust
Register

The OSCB Frequency Adjust register is used to adjust the frequency in case of external components inaccuracies. It allows a frequency variation about $15 \%$ around 12 MHz with a step of around $1 \%$.

Table 4. OSCBFA Register OSCBFA- Oscillator B Frequency Adjust Register (9Fh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| OSCBFA7 | OSCBFA6 | OSCBFA5 | OSCBFA4 | OSCBFA3 | OSCBFA2 | OSCBFA1 | OSCBFAO |
| Bit <br> Number | Bit Mnemonic | Description |  |  |  |  |  |
| 7-0 | $\begin{gathered} \text { OSCBFA } \\ 7-0 \end{gathered}$ | OSCB Frequency adjust <br> The reset value to have 12 MHz is 01110110 . It is possible to modify this value in order to increase or decrease the frequency. |  |  |  |  |  |

Reset Value = 0111 0110b
Not bit addressable

## Clock Selection Register

The clock selection register is used to define the clock system behavior.
Table 5. CKSEL Register
CKSEL - Clock Selection Register (85h)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | - | CKS1 | CKSO |
| $\begin{gathered} \text { Bit } \\ \text { Number } \end{gathered}$ | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |  |  |  |  |  |
| 6 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |  |  |  |  |  |
| 5 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |  |  |  |  |  |
| 4 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |  |  |  |  |  |
| 3 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |  |  |  |  |  |
| 2 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |  |  |  |  |  |
| 1 | CKS1 | Active Clock Selector 1and Active Clock Selector 0 <br> These bits are used to select the active oscillator <br> 11: The crystal oscillator is selected <br> 10: The high accuracy RC oscillator is selected <br> 01: The low power consumption RC oscillator is selected <br> 00: Reserved |  |  |  |  |  |
| 0 | CKSO |  |  |  |  |  |  |

Reset Value = XXXX XX"RST_OSC1" "RST_OSC0" b
Not bit addressable

## Clock Prescaler Register

## Clock Control Register

This register is used to reload the clock prescaler of the CPU and peripheral clock.

Table 6. CKRL Register
CKRL - Clock prescaler Register (97h)


Reset Value = XXXX 1000b
Not bit addressable
This register is used to control the X2 mode of the CPU and peripheral clock.

Table 7. CKCON Register
CKCON - Clock Control Register (8Fh)


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :--- | \left\lvert\, | $7-1$ | - |
| :---: | :---: |
| Reserved |  |
| 0 | X2 | | X2 Mode |
| :--- |
| Set to force X2 mode (CkOut $=$ OscOut) |
| Clear to use the prescaler mode (CkOut $=$ OscOut / (2*(16-M))) |\right.

Reset Value $=0000$ 0000b
Not bit addressable

## Power Modes

## Overview

## Operating Modes

## Normal Mode

## Idle Mode

Entering Idle Mode

As seen in the previous chapter it is possible to modify the AT8xEB5114 clock management in order to have less consumption.
For applications where power consumption is a critical factor, three power modes are provided:

- Normal (running) mode
- Idle mode
- Power-down mode

In order to increase ADC accuracy, a Quiet mode also exits. This mode is a pseudo idle mode in which the CPU and all the peripherals except the AD converter are disabled.
Power modes are controlled by PCON SFR register.
Table 8 summarizes all the power modes and states that AT8xEB5114 can encounter. It shows which parts of AT8xEB5114 are running depending on the operating mode.

Table 8. Operating Modes

| Operating Mode | Prescaler | Oscillator | POR | CPU | Peripherals |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Power Down |  |  | X |  |  |
| Under Reset |  | A, B or C | X |  |  |
| Start | X | A, B or C | X | X |  |
| Running | (X) | A, B or C | X | X | X |
| Idle | (X) | A, B or C | X |  | X |
| Quiet | (X) | A, B or C | X |  | only ADC |

In normal mode, the oscillator, the CPU and the peripherals are running. The prescaler can also be activated.

- The CPU and the peripherals clock depends on the software selection using CKCON, OSCCON, CKSEL and CKRL registers
- CKS bits select either OSCA, OSCB, or OSCC
- CKRL register determines the frequency of the selected clock, unless X2 bit is set. In this case the prescaler/divider is not used, so CPU core needs only 6-clock periods per machine cycle.
It is always possible to switch dynamically by software from one to another oscillator by changing CKS bits, a synchronization cell allows to avoid any spike during transition.

The idle mode allows to reduce consumption by freezing the CPU. All the peripherals continue running.

An instruction that sets PCON. 0 causes that to be the last instruction executed before going into Idle mode.
In Idle mode, the internal clock signal is gated off to the CPU, but not to the interrupt, and the peripheral functions. The CPU status is entirely preserved: the Stack Pointer, Program Counter, Program Status Word, Accumulator and all other registers maintain

Exit from Idle Mode

Quiet Mode

Power-down Mode

Entering Power-down Mode

Exit from Power-down Mode
their data during Idle. The port pins hold the logical states they had at the time Idle was activated. ALE and PSEN are held at logic high levels. The different operating modes are summarized on Table 10 on page 21.

There are two ways to terminate idle mode. Activation of any enabled interrupt will cause PCON. 0 to be cleared by hardware, terminating Idle mode. The interrupt will be serviced, and following RETI the next instruction to be executed will be the one following the instruction that put the device into idle. Exit from idle mode will leave the oscillators control bits on OSCON and CKS registers unchanged.

The flag bits GF0 and GF1 can be used to give an indication if an interrupt occurred during normal operation or during an Idle mode. For example, an instruction that activates Idle mode can also set one or both flag bits. When Idle is terminated by an interrupt, the interrupt service routine can examine the flag bits.
The other way of terminating the Idle mode is with a hardware reset. Since the clock oscillator is still running, the hardware reset needs to be held active for only two machine cycles (24 oscillator periods) to complete the reset.

In both cases, PCON. 0 is cleared by hardware.
The quiet mode is a pseudo idle mode in which the CPU and all the peripherals except the AD converter are down. For more details, See "Analog-to-Digital Converter (ADC)" on page 57.

To save maximum power, a power-down mode can be invoked by software (refer to Table 11 on page 22). In power-down mode, the oscillator is stopped and the instruction that invoked power-down mode is the last instruction executed. The internal RAM and SFRs retain their value until the power-down mode is terminated. $\mathrm{V}_{\mathrm{CC}}$ can be lowered to save further power.

An instruction that sets PCON. 1 causes that to be the last instruction executed before going into the power-down mode.
The ports status under power-down is the previous status before entering this power mode.

Either a hardware reset or an external interrupt (low level) on INT0 or INT1 (if enabled) can cause an exit from power-down. To properly terminate power-down, the reset or external interrupt should not be executed before $\mathrm{V}_{\mathrm{CC}}$ is restored to its normal operating level and must be held active long enough for the oscillator to restart and stabilize.

Exit from power-down by external interrupt does not affect the SFRs and the internal RAM content.

Figure 6. Power-down Exit Waveform


By a hardware Reset, the CPU will restart in the mode defined by the RST_OSC1 and RST_OSCO bits in HSB.

By INT1 and INT0 interruptions (if enabled), the oscillators control bits on OSCON and CKSEL will be kept, so the selected oscillator before entering in power-down mode will be activated. Only external interrupts INT0 and INT1 are useful to exit from power-down.
Note: Exit from power down mode doesn't depend on ITO and IT1 configurations. It is only possible to exit from power down mode on a low level on these pins.

Holding the pin low restarts the oscillator but bringing the pin high completes the exit as detailed in Figure 6. When both interrupts are enabled, the oscillator restarts as soon as one of the two inputs is held low and power down exit will be completed when the first input is released. In this case the higher priority interrupt service routine is executed.

Table 9 shows the state of ports during idle and power-down modes.

Table 9. Ports State

| Mode | Program Memory | Port3 | Port4 |
| :---: | :---: | :---: | :---: |
| Idle | Internal | Port Data | Port Data |
| Power Down | Internal | Port Data | Port Data |

Table 10. Operating Modes

| PD | IDLE | CKS1 | CKS0 | OSCCEN | OSCBEN | OSCAEN | Selected Mode | Comment |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :--- | :--- |
| 0 | 0 | 1 | 1 | X | X | 1 | NORMAL MODE A | OSCA: XTAL clock |
| X | X | 1 | 1 | X | X | 0 | INVALID | no active clock |
| 0 | 0 | 1 | 0 | X | 1 | X | NORMAL MODE B, | OSCB: high accuracy RC clock |
| X | X | 1 | 0 | X | 0 | X | INVALID | no active clock |
| 0 | 0 | 0 | 1 | 1 | X | X | NORMAL MODE C, | OSCC: low consumption RC clock |
| X | X | 0 | 1 | 0 | X | X | INVALID | no active clock |
| 0 | 1 | 1 | 1 | X | X | 1 | IDLE MODE A | The CPU is off, OSCA supplies the <br> peripherals |
| 0 | 1 | 1 | 0 | X | 1 | X | IDLE MODE B | The CPU is off, OSCB supplies the <br> peripherals |
| 0 | 1 | 0 | 1 | 1 | X | X | IDLE MODE C | The CPU is off, OSCC supplies the <br> peripherals |
| 1 | X | X | X | X | X | X | POWER DOWN | The CPU is off, OSCA, OSCB and <br> OSCC are stopped |

Power Modes Control Registers

Table 11. PCON Register
PCON (S:87h)
Power configuration Register

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | GF1 | GF0 | PD | IDL |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :--- | :--- |
| 7 |  | Reserved |
| 6 | GF1 | Reserved <br> 5 |
| 4 | Get and Cleared by user for general purpose usage. |  |
| 3 | PD | General Purpose flag 0 <br> Set and Cleared by user for general purpose usage. |
| 2 | Power-down Mode bit <br> Cleared by hardware when an interrupt or reset occurs. <br> Set to activate the Power-down mode. <br> If IDL and PD are both set, PD takes precedence. |  |
| 1 | IDL | Idle Mode bit <br> Cleared by hardware when an interrupt or reset occurs. <br> Set to activate the Idle mode. <br> If IDL and PD are both set, PD takes precedence. |
| 0 |  |  |

Reset Value $=00 \mathrm{XX}$ XX00b

## AUXR Register

Table 12. AUXR Register
AUXR - Auxiliary Register (8Eh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DPU | - | - | LOWVD | - | - | - | - |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | DPU | Disable Pull up <br> Set to disable each pull up on all ports. <br> Clear to connect all pull-ups on each port. |
| 6 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 5 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 4 | LOWVD | Low Voltage Detection <br> This bit is clear by hardware when the supply voltage is under Vpfdp value. This bit is set by hardware as soon the supply voltage is greater than Vpfdp value. |
| 3-1 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 0 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |

Reset Value = 0XX0 XXXXb
Not bit addressable

## Timers/Counters

## Introduction

## Timer/Counter Operations

## Timer 0

The AT8xEB5114 implements two general-purpose, 16-bit Timers/Counters. Although they are identified as Timer 0, Timer 1, they can be independently configured each to operate in a variety of modes as a Timer or as an event Counter. When operating as a Timer, a Timer/Counter runs for a programmed length of time, then issues an interrupt request. When operating as a Counter, a Timer/Counter counts negative transitions on an external pin. After a preset number of counts, the Counter issues an interrupt request.
The Timer registers and associated control registers are implemented as addressable Special Function Registers (SFRs). Two of the SFRs provide programmable control of the Timers as follows:

- Timer/Counter mode control register (TMOD) and Timer/Counter control register (TCON) control both Timer 0 and Timer 1.

The various operating modes of each Timer/Counter are described below.
A basic operation is Timer registers THx and TLx $(x=0,1)$ connected in cascade to form a 16-bit Timer. Setting the run control bit (TRx) in the TCON register (see Figure 15) turns the Timer on by allowing the selected input to increment TLx. When TLx overflows it increments THx and when THx overflows it sets the Timer overflow flag (TFx) in the TCON register. Setting the TRx does not clear the THx and TLx Timer registers. Timer registers can be accessed to obtain the current count or to enter preset values. They can be read at any time but the TRx bit must be cleared to preset their values, otherwise the behavior of the Timer/Counter is unpredictable.

The C/Tx\# control bit selects Timer operation or Counter operation by selecting the divided-down system clock or the external pin Tx as the source for the counted signal. The TRx bit must be cleared when changing the operating mode, otherwise the behavior of the Timer/Counter is unpredictable.
For Timer operation (C/Tx\# = 0), the Timer register counts the divided-down system clock. The Timer register is incremented once every peripheral cycle.
For Counter operation (C/Tx\# = 1), the Timer register counts the negative transitions on the external input pin Tx. The external input is sampled during every S5P2 state. The Programmer's Guide describes the notation for the states in a peripheral cycle. When the sample is high in one cycle and low in the next one, the Counter is incremented. The new count value appears in the register during the next S3P1 state after the transition has been detected. Since it takes 12 states ( 24 oscillator periods in X1 mode) to recognize a negative transition, the maximum count rate is $1 / 24$ of the oscillator frequency in X1 mode. There are no restrictions on the duty cycle of the external input signal, but to ensure that a given level is sampled at least once before it changes, it should be held for at least one full peripheral cycle.

Timer 0 functions as either a Timer or an event Counter in four operating modes. Figure 7 to Figure 10 show the logic configuration of each mode.
Timer 0 is controlled by the four lower bits of the TMOD register (see Figure 16) and bits $0,1,4$ and 5 of the TCON register (see Figure 15). The TMOD register selects the method of Timer gating (GATE0), Timer or Counter operation (T/CO\#) and the operating mode (M10 and M00). The TCON register provides Timer 0 control functions: overflow flag (TFO), run control bit (TRO), interrupt flag (IE0) and interrupt type control bit (ITO).
For normal Timer operation (GATE0=0), setting TR0 allows TLO to be incremented by the selected input. Setting GATE0 and TR0 allows external pin INTO\# to control Timer
operation.
Timer 0 overflow (count rolls over from all 1s to all 0s) sets the TF0 flag and generates an interrupt request.
It is important to stop the Timer/Counter before changing modes.
Mode 0 configures Timer 0 as a 13-bit Timer which is set up as an 8-bit Timer (TH0 register) with a modulo-32 prescaler implemented with the lower five bits of the TL0 register (see Figure 7). The upper three bits of the TLO register are indeterminate and should be ignored. Prescaler overflow increments the TH0 register.

Figure 7. Timer/Counter $x(x=0$ or 1$)$ in Mode 0


Mode 1 (16-bit Timer) Mode 1 configures Timer 0 as a 16-bit Timer with the TH0 and TL0 registers connected in a cascade (see Figure 8). The selected input increments the TLO register.

Figure 8. Timer/Counter $x(x=0$ or 1$)$ in Mode 1


Mode 2 (8-bit Timer with Auto- Mode 2 configures Timer 0 as an 8 -bit Timer (TLO register) that automatically reloads Reload) from the TH0 register on overflow (see Figure 9). TLO overflow sets the TFO flag in the TCON register and reloads TLO with the contents of THO, which is preset by the software. When the interrupt request is serviced, the hardware clears TFO. The reload leaves THO unchanged. The next reload value may be changed at any time by writing it to the TH0 register.

Figure 9. Timer/Counter $\mathrm{x}(\mathrm{x}=0$ or 1 ) in Mode 2


Mode 3 (Two 8-bit Timers)
Mode 3 configures Timer 0 so that registers TLO and TH0 operate as 8 -bit Timers (see Figure 10). This mode is provided for applications requiring an additional 8 -bit Timer or Counter. TLO uses the Timer 0 control bits C/TO\# and GATEO in the TMOD register, and TR0 and TF0 in the TCON register in the normal manner. TH0 is locked into a Timer function (counting $\mathrm{F}_{\text {UART }}$ ) and takes over use of the Timer 1 interrupt (TF1) and run control (TR1) bits. Thus, operation of Timer 1 is restricted when Timer 0 is in mode 3.

Figure 10. Timer/Counter 0 in Mode 3: Two 8-bit Counters


Timer 1
Timer 1 is identical to Timer 0 except for Mode 3 which is a hold-count mode. The following comments help to understand the differences:

- Timer 1 functions as either a Timer or an event Counter in the three operating modes. Figure 7 to Figure 9 show the logical configuration for modes 0,1 , and 2. Mode 3 of Timer 1 is a hold-count mode.
- Timer 1 is controlled by the four high-order bits of the TMOD register (see Figure 16) and bits 2, 3, 6 and 7 of the TCON register (see Figure 15). The TMOD register selects the method of Timer gating (GATE1), Timer or Counter operation (C/T1\#) and the operating mode (M11 and M01). The TCON register provides Timer 1 control functions: overflow flag (TF1), run control bit (TR1), interrupt flag (IE1) and the interrupt type control bit (IT1).
- Timer 1 can serve as the Baud Rate Generator for the Serial Port. Mode 2 is best suited for this purpose.

Mode 0 (13-bit Timer)

Mode 1 (16-bit Timer)

Mode 2 (8-bit Timer with AutoReload)

Mode 3 (Halt)

- For normal Timer operation (GATE1=0), setting TR1 allows TL1 to be incremented by the selected input. Setting GATE1 and TR1 allows external pin INT1\# to control Timer operation.
- Timer 1 overflow (count rolls over from all 1 s to all 0s) sets the TF1 flag and generates an interrupt request.
- When Timer 0 is in mode 3, it uses Timer 1's overflow flag (TF1) and run control bit (TR1). For this situation, use Timer 1 only for applications that do not require an interrupt (such as a Baud Rate Generator for the Serial Port) and switch Timer 1 in and out of mode 3 to turn it off and on.
- It is important to stop the Timer/Counter before changing modes.

Mode 0 configures Timer 1 as a 13-bit Timer, which is set up as an 8-bit Timer (TH1 register) with a modulo-32 prescaler implemented with the lower 5 bits of the TL1 register (see Figure 7). The upper 3 bits of TL1 register are indeterminate and should be ignored. Prescaler overflow increments the TH1 register.

Mode 1 configures Timer 1 as a 16-bit Timer with TH1 and TL1 registers connected in cascade (see Figure 8). The selected input increments the TL1 register.

Mode 2 configures Timer 1 as an 8-bit Timer (TL1 register) with automatic reload from the TH1 register on overflow (see Figure 9). TL1 overflow sets the TF1 flag in the TCON register and reloads TL1 with the contents of TH1, which is preset by the software. The reload leaves TH 1 unchanged.

Placing Timer 1 in mode 3 causes it to halt and hold its count. This can be used to halt Timer 1 when the TR1 run control bit is not available i.e. when Timer 0 is in mode 3.

Registers
Table 13. TCON (S:88h)
Timer/Counter Control Register

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| TF1 | TR1 | TF0 | TR0 | IE1 | IT1 | IEO | ITO |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7 | TF1 | Timer 1 Overflow flag <br> Cleared by the hardware when processor vectors to interrupt routine. Set by the hardware on Timer 1 register overflows. |  |  |  |  |  |
| 6 | TR1 | Timer 1 Run Control bit <br> Clear to turn off Timer/Counter 1. Set to turn on Timer/Counter 1. |  |  |  |  |  |
| 5 | TF0 | Timer 0 Overflow flag <br> Cleared by the hardware when processor vectors to interrupt routine. Set by the hardware on Timer 0 register overflows. |  |  |  |  |  |
| 4 | TR0 | Timer 0 Run Control bit <br> Clear to turn off Timer/Counter 0. Set to turn on Timer/Counter 0. |  |  |  |  |  |
| 3 | IE1 | Interrupt 1 Edge flag <br> Cleared by the hardware as soon as the interrupt is processed. Set by the hardware when external interrupt is detected on the $\overline{\mathrm{NT} 1} \mathrm{pin}$. |  |  |  |  |  |
| 2 | IT1 | Interrupt 1 Type Control bit <br> Clear to select low level active for external interrupt 1 (INT1). <br> Set to select sensitive edge trigger for external interrupt 1. The sensitive edge (Rising or Falling) is determined by ESB1 value (Edge Selection Bit 1) in IOR (Interrupt Option Register). |  |  |  |  |  |
| 1 | IE0 | Interrupt 0 Edge flag <br> Cleared by the hardware as soon as the interrupt is processed. Set by the hardware when external interrupt is detected on $\overline{\mathrm{NTO}}$ pin. |  |  |  |  |  |
| 0 | IT0 | Interrupt 0 Type Control bit <br> Clear to select low level active trigger for external interrupt 0 (INTO). Set to select sensitive edge trigger for external interrupt 0 . The sensitive edge (Rising or Falling) is determined by ESB0 (Edge Selection Bit 0) in IOR (Interrupt Option Register). |  |  |  |  |  |

Reset Value $=0000$ 0000b

Table 14. IOR (S:A5h)
Interrupt Option Register.

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | - | ESB1 | ESB0 |


| Bit Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :---: |
| 7-2 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 1 | ESB1 | Edge Selection bit for $\overline{\mathrm{NT} 1}$ <br> Clear to select falling edge sensitive for $\overline{\mathrm{NT} 1}$ pin. Set to select rising edge sensitive for $\overline{\mathrm{NNT}}$ pin. |
| 0 | ESB0 | Edge Selection bit for $\overline{\text { INTO }}$ <br> Clear to select falling edge sensitive for $\overline{\mathrm{NTO}}$ pin. <br> Set to select rising edge sensitive for $\overline{\mathrm{NTO}}$ pin. |

Reset Value $=$ XXXX XX00b

Table 15. TMOD Register
TMOD (S:89h)
Timer/Counter Mode Control Register.

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| GATE1 | C/T1\# | M11 | M01 | GATE0 | C/TO\# | M10 | M00 |


| $\begin{gathered} \text { Bit } \\ \text { Number } \end{gathered}$ | Bit Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | GATE1 | Timer 1 Gating Control bit <br> Clear to enable Timer counter 1 whenever TR1 bit is set. <br> Set to enable Timer counter 1 only while INT1\# pin is high and TR1 bit is set. |
| 6 | C/T1\# | Timer 1 Counter/Timer Select bit <br> Clear for Timer operation: Timer 1 counts the divided-down system clock. <br> Set for Counter operation: Timer 1 counts negative transitions on external pin T1. |
| 5 | M11 | Timer 1 Mode Select bitsM11 M01 Operating mode  <br> 0 0 Mode 0: 8 -bit Timer/Counter (TH1) with 5-bit prescaler (TL1). <br> 0 1 Mode 1: 16-bit Timer/Counter. <br> 1 0 Mode 2: 8 -bit auto-reload Timer/Counter (TL1). Reloaded from <br>    TH1 at overflow. <br> 1   Mode 3:Timer 1 halted. Retains count. |
| 4 | M01 |  |
| 3 | GATE0 | Timer 0 Gating Control bit <br> Clear to enable Timer counter 0 whenever TR0 bit is set. Set to enable Timer counter 0 only while INTO\# pin is high and TRO bit is set. |
| 2 | C/T0\# | Timer 0 Counter/Timer Select bit <br> Clear for Timer operation: Timer 0 counts the divided-down system clock. <br> Set for Counter operation: Timer 0 counts negative transitions on external pin T0. |
| 1 | M10 |  |
| 0 | M00 |  |

Reset Value $=0000$ 0000b

Table 16. THO Register
TH0 (S:8Ch)
Timer 0 High Byte Register.

| 7 | 6 | 5 | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | 0 |
| :---: | :---: | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7:0 |  | High Byte of Timer 0. |  |  |  |  |  |

Reset Value $=0000$ 0000b

Table 17. TLO Register
TLO (S:8Ah)
Timer 0 Low Byte Register.

| 7 | 6 | 5 | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | 1 |  |
| :---: | :---: | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| $7: 0$ |  | Low Byte of Timer 0. |  |  |  |  |  |

Reset Value $=0000$ 0000b

Table 18. TH1 Register
TH1 (S:8Dh)
Timer 1 High Byte Register.

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :--- | :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |  |  |  |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| $7: 0$ |  | High Byte of Timer 1. |  |  |  |  |  |

Reset Value $=00000000 \mathrm{~b}$

Table 19. TL1 Register

## TL1 (S:8Bh)

Timer 1 Low Byte Register.

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Bit | Description |  |  |  |  |  |
| 7:0 |  | Low Byte of Timer 1. |  |  |  |  |  |

Reset Value $=0000$ 0000b

## Ports

The AT8xEB5114 has 2 I/O ports, port 3, and port 4.
All port3 and port4 I/O port pins on the AT8xEB5114 may be software configured to one of four types on a bit-by-bit basis, as shown below in Table 20. These are: quasi-bidirectional (standard 80C51 port outputs), push-pull, open drain, and input only. Two configuration registers for each port select the output type for each port pin.

Table 20. Port Output Configuration setting using PxM1 and PxM2 registers ( $3 \leq x \leq 4$ )

| PxM1.(2y+1) bit | PxM1.(2y) bit | $(\mathbf{0} \leq \mathbf{y} \leq 3)$ Port Output Mode |
| :---: | :---: | :--- |
| 0 | 0 | Quasi bidirectional |
| 0 | 1 | Push-Pull |
| 1 | 0 | Input Only (High Impedance) |
| 1 | 1 | Open Drain |


| PxM2.(2y-7) bit | PxM2.(2y-8) bit | (4 $\leq y \leq 7)$ Port Output Mode |
| :---: | :---: | :--- |
| 0 | 0 | Quasi bidirectional |
| 0 | 1 | Push-Pull |
| 1 | 0 | Input Only (High Impedance) |
| 1 | 1 | Open Drain |

## Port Types

## Quasi-Bidirectional Output Configuration

The default port output configuration for standard AT8xEB5114 I/O ports is the quasibidirectional output that is common on the 80C51 and most of its derivatives. This output type can be used as both an input and output without the need of reconfiguring the port. This is possible because when the port outputs a logic high, it is weakly driven, allowing an external device to pull the pin low. When the pin is pulled low, it is driven strongly and able to sink a fairly large current. These features are somewhat similar to an open drain output except that there are three pull-up transistors in the quasi-bidirectional output that serve different purposes. One of these pull-ups, called the "weak" pull-up, is turned on whenever the port latch for the pin contains a logic 1 . The weak pull-up sources a very small current that will pull the pin high if it is left floating. A second pull-up, called the "medium" pull-up, is turned on when the port latch for the pin contains a logic 1 and the pin itself is also at a logic 1 level. This pull-up provides the primary source current for a quasi-bidirectional pin that is outputting a 1 . If a pin that has a logic 1 on it is pulled low by an external device, the medium pull-up turns off, and only the weak pull-up remains on. In order to pull the pin low under these conditions, the external device has to sink enough current to overpower the medium pull-up and take the voltage on the port pin below its input threshold.

The third pull-up is referred to as the "strong" pull-up. This pull-up is used to speed up low-to-high transitions on a quasi-bidirectional port pin when the port latch changes from a logic 0 to a logic 1. When this occurs, the strong pull-up turns on for a brief time, two CPU clocks, in order to pull the port pin high quickly. Then it turns off again.
The quasi-bidirectional port configuration is shown in Figure 11.

Figure 11. Quasi-Bidirectional Output


## Open Drain Output Configuration

The open drain output configuration turns off all pull-ups and only drives the pull-down transistor of the port driver when the port latch contains a logic 0 . To be used as a logic output, a port configured in this manner must have an external pull-up, typically a resistor tied to $\mathrm{V}_{\mathrm{DD}}$. The pull-down for this mode is the same as the quasi-bidirectional mode. The open drain port configuration is shown in Figure 12.

Figure 12. Open Drain Output


## Push-Pull Output Configuration

The push-pull output configuration has the same pull-down structure as both the open drain and the quasi-bidirectional output modes, but provides a continuous strong pull-up when the port latch contains a logic 1. The push-pull mode may be used when more source current is needed from a port output. The push-pull port configuration is shown in Figure 13.

Figure 13. Push-Pull Output


Input only Configuration
The input only configuration is a pure input with neither pull-up nor pull-down. The input only configuration is shown in Figure 13.

Figure 14. Input only


## Ports Description

Ports P3 and P4
The inputs of each I/O port of the AT8xEB5114 are TTL level Schmitt triggers with hysteresis.

Table 21. P3M1 Register
P3M1 Address (D5h)

| P3M1.7 | P3M1.6 | P3M1.5 | P3M1.4 | P3M1.3 | P3M1. 2 | P3M1.1 | P3M1.0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{gathered} \text { Bit } \\ \text { Number } \end{gathered}$ | Bit Mnemonic | Description |  |  |  |  |  |
| 7-6 | P3M1.7-6 | Port 3.3 Output configuration bit See Table 20 for configuration definition |  |  |  |  |  |
| 5-4 | P3M1.5-4 | Port 3.2 Output configuration bit See Table 20 for configuration definition |  |  |  |  |  |
| 3-2 | P3M1.3-2 | Port 3.1 Output configuration bit See Table 20 for configuration definition |  |  |  |  |  |
| 1-0 | P3M1.1-0 | Port 3.0 Output configuration bit SeeTable 20 for configuration definition |  |  |  |  |  |

Reset value $=00000000$

Table 22. P3M2 Register
P3M2 Address (E4h)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| P3M2.7 | P3M2.6 | P3M2.5 | P3M2.4 | P3M2.3 | P3M2.2 | P3M2.1 | P3M2.0 |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :--- | :--- |
| $7-6$ | P3M2.7-6 | Port 3.7 Output configuration bit <br> SeeTable 20 for configuration definition |
| $5-4$ | P3M2.5-4 | Port 3.6 Output configuration bit <br> See Table 20 for configuration definition |
| $3-2$ | P3M2.3-2 | Port 3.5 Output configuration bit <br> See Table 20 for configuration definition |
| $1-0$ | P3M2.1-0 | Port 3.4 Output configuration bit <br> See Table 20 for configuration definition |

Reset value $=00000000$

Table 23. P4M1 Register
P4M1 Address (D6h)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| P4M1.7 | P4M1.6 | P4M1.5 | P4M1.4 | P4M1.3 | P4M1.2 | P4M1.1 | P4M1.0 |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :--- | :--- |
| $7-6$ | P4M1.7-6 | Port 4.3 Output configuration bit <br> See Table 20 for configuration definition |
| $5-4$ | P4M1.5-4 | Port 4.2 Output configuration bit <br> See Table 20 for configuration definition |
| $3-2$ | P4M1.3-2 | Port 4.1 Output configuration bit <br> See Table 20 for configuration definition |
| $1-0$ | P4M1.1-0 | Port 4.0 Output configuration bit <br> See Table 20 for configuration definition |

Reset value $=00000000$

Dual Data Pointer Register (DDPTR)

The additional data pointer can be used to speed up code execution and reduce code size in a number of ways.
The dual DPTR structure is a way by which the chip will specify the address of an external data memory location. There are two 16-bit DPTR registers that address the external memory, and a single bit called DPS = AUXR1/bit0 (See Figure 15) that allows the program code to switch between them.

Figure 15. Use of Dual Pointer


Table 24. AUXR1: Auxiliary Register 1

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | 0 | - | DPS |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 6 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 5 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 4 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 3 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 2 | 0 | Reserved always stuck at 0 |
| 1 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 0 | DPS | Data Pointer Selection <br> Clear to select DPTR0. Set to select DPTR1. |

Note: User software should not write 1's to reserved bits. These bits may be used in future 8051 family products to invoke new feature. In that case, the reset value of the new bit will be 0 , and its active value will be 1 . The value read from a reserved bit is indeterminate.

## Application

Software can take advantage of the additional data pointers to both increase speed and reduce code size, for example, block operations (copy, compare, search...) are well served by using one data pointer as a 'source' pointer and the other one as a 'destination' pointer.

## ASSEMBLY LANGUAGE

```
; Block move using dual data pointers
; Destroys DPTRO, DPTR1, A and PSW
; note: DPS exits opposite of entry state
; unless an extra INC AUXR1 is added
OOA2 AUXR1 EQU OA2H
;
0000 909000MOV DPTR,#SOURCE ; address of SOURCE
0003 05A2 INC AUXR1 ; switch data pointers
0005 90A000 MOV DPTR,#DEST ; address of DEST
0008 LOOP:
0008 05A2 INC AUXR1 ; switch data pointers
OOOA EO MOVX A,@DPTR ; get a byte from SOURCE
OOOB A3 INC DPTR ; increment SOURCE address
OOOC 05A2 INC AUXR1 ; switch data pointers
OOOE FO MOVX @DPTR,A ; write the byte to DEST
OOOF A3 INC DPTR ; increment DEST address
OO10 70F6JNZ LOOP ; check for O terminator
0 0 1 2 ~ 0 5 A 2 ~ I N C ~ A U X R 1 ~ ; ~ ( o p t i o n a l ) ~ r e s t o r e ~ D P S ~
```

INC is a short (2 bytes) and fast (12 clocks) way to manipulate the DPS bit in the AUXR1 SFR. However, note that the INC instruction does not directly force the DPS bit to a particular state, but simply toggles it. In simple routines, such as the block move example, only the fact that DPS is toggled in the proper sequence matters, not its actual value. In other words, the block move routine works the same whether DPS is ' 0 ' or ' 1 ' on entry. Observe that without the last instruction (INC AUXR1), the routine will exit with DPS in the opposite state.

PWM Unit 0 (PWMU0)
The PWM unit 0 allows to generate precise pulse width modulation with variable duty cycle and frequency.
The PWMUO consists on a dedicated 16 bits auto reload counter/timer which serves as a time base for the generation of 3 independent PWM signals.
Its clock input can be programmed to count any one of the following signals:

- Peripheral clock, CkIdle
- Timer 0 overflow
- External input on W0CI (P4.0)

The PWMU0 timer/counter shares several external I/O. These pins are listed below. If a port is not used for the PWMU0, it can still be used for standard I/O.

| PWMU0 Component | External I/O Pin |
| :---: | :---: |
| 16-bit Counter | WOCI (P4.0) |
| 16-bit Module 0 | WOM0 (P3.0) |
| 16-bit Module 1 | WOM1 (P3.1) |
| 16-bit Module 2 | WOM2 (P3.3) |

PWMUO Timer
The PWMU0 timer is a common 16 bits time base for all three modules (See Figure 16). The timer count source is determined from the W0CPS1 and WOCPS0 bits in the WOMOD register (See Table 26) and can be programmed to run at:

- Peripheral clock, Ckldle
- Timer 0 overflow
- External input on WOCI (P1.2)

The output frequency depends on the timer source and also on the WOF Registers. Indeed, the timer/counter counts from zero up to a value loaded via SWOF registers. Each time the counter is higher or equal to the SWOF shadow registers value, WOC registers are automatically reloaded with zero. If the WOUP bit is set, the shadow SW0F registers are reloaded with the contents of WOF registers when the WOC overtakes. This prevents frequency drift (See Figure 16.).
Note: If the PWMUO is Off (WOR bit in WOCON not set), the contents of WOFH and WOFL are automatically copied on the shadow registers SWOFH and SWOFL. This allows to charge the correct comparison values in order to have the wanted frequency as soon as the PWM is turned on.

Figure 16. PWMU0 Timer/Counter


Table 25. WOCON: PWMU0 Control register WOCON - PWMU0 Control Register (E8h)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WOUP | WOR | - |  | woos | WOEN2 | W0EN1 | WOENO |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7 | W0UP | PWMUO update bit <br> Set by software to request the load of all shadow registers on the next overtaking of the WOC counter. Reset by hardware after the loading of the shadow registers. |  |  |  |  |  |
| 6 | WOR | PWMUO Run control bit <br> Set by software to turn the PWMU0 counter on. Must be cleared by software to turn the PWMU0 counter off. |  |  |  |  |  |
| 5-4 | - | Not used |  |  |  |  |  |
| 3 | W0OS | Pin W0M1 PWMU0 Output Selection <br> 0 WOM1 is PWM module 1 XOR PWM module2 output 1 W0M1 is PWM module 1 output |  |  |  |  |  |
| 2 | W0EN2 | PWMUO Module 2 enable bit Enable PWMUO module 2 if set. |  |  |  |  |  |
| 1 | W0EN1 | PWMUO Module 1 enable bit Enable PWMU0 module 1if set. |  |  |  |  |  |
| 0 | W0EN0 | PWMUO Module 0 enable bit Enable PWMUO module 0 if set. |  |  |  |  |  |

Reset Value $=00 \mathrm{XX}$ 0000b
Bit addressable

Table 26. WOMOD: PWMUO Counter Mode Register WOMOD - PWMUO Counter Mode Register (E9h)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WOCPS1 | WOCPS0 | - | - | - | WOINV2 | WOINV1 | WOINV0 |


| $\begin{gathered} \text { Bit } \\ \text { Number } \end{gathered}$ | Bit Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | W0CPS1 | PWMU0 Count Pulse Select bit1 |
| 6 | WOCPSO | PWMUO Count Pulse Select bit0 <br> CPS1 CPS0 Selected PWMU0 input <br> 00 Internal clock $\mathrm{f}_{\text {Ckldle }}$ <br> 01 Reserved <br> 10 Timer 0 Overflow <br> 11 External clock input on WOCl at max rate $=\mathrm{f}_{\text {Cklde }} / 4$ |
| 5-3 | - | Not used |
| 2 | W0INV2 | PWMUO Module 2 inverter bit <br> Select the output PWM mode. If set, PWM module 2 output starts with high level. |
| 1 | W0INV1 | PWMUO Module 1 inverter bit <br> Select the output PWM mode. If set, PWM module 1 output starts with high level. |
| 0 | W0INV0 | PWMUO Module 0 inverter bit <br> Select the output PWM mode. If set, PWM module 0 output starts with high level. |

Reset Value $=00 \mathrm{XX}$ X000b
Not bit addressable
Because they use the same timer, all three modules have the same frequency determined by the shadow SWOF registers.

Table 27. WOFH: PWMUO frequency high control register WOFH - PWMUO Frequency Control Register (EAh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W0F15 | W0F14 | W0F13 | W0F12 | W0F11 | W0F10 | W0F9 | W0F8 |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| $7-0$ | W0F15-8 | PWMU0 high bits counter control frequency <br> The PWMU0 counter is counting from zero up to W1F15-0 value. |  |  |  |  |  |

Reset Value = 1111 1111b
Not bit addressable

Table 28. WOFL: PWMU0 frequency low control register WOFL - PWMU0 Frequency Control Register (EBh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W0F7 | W0F6 | W0F5 | W0F4 | W0F3 | W0F2 | W0F1 | W0FO |
| Bit <br> Number | Bit Mnemonic | Description |  |  |  |  |  |
| 7-0 | W0F7-0 | PWMUO low bits counter control frequency <br> The PWMU0 counter is counting from zero up to WOF15-0 value. |  |  |  |  |  |

Reset Value $=1111$ 1111b
Not bit addressable

Table 29. W0CH: PWMU0 counter high control register W0CH - PWMU0 Counter Control Register (ECh)

| 7 | 6 | 5 | 4 | 0 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W0C15 | W0C14 | W0C13 | W0C12 | W0C11 | W0C10 | W0C9 | W0C8 |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| $7-0$ | W0C15-8 | PWMU0 high bits counter frequency. |  |  |  |  |  |

Reset Value $=0000$ 0000b
Not bit addressable

Table 30. WOCL: PWMU0 counter low control register W0CL - PWMU0 Counter Control Register (EDh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W0C7 | W0C6 | W0C5 | W0C4 | W0C3 | W0C2 | W0C1 | W0C0 |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| $7-0$ | W0C7-0 | PWMU0 low bits counter frequency. |  |  |  |  |  |

Reset Value $=0000$ 0000b
Not bit addressable

PWMUO Output Generation

All the PWMUO modules have the same frequency determined by the WOF register. But each module has its own duty cycle determined by the WORn Register. ( n is the module number).

When the WOC content is lower than the value programmed via the W0Rn registers, the output is the WOINVn-bit (low if 0 , high if 1 ). When it is equal or higher, the output is the opposite of this WOINVn-bit (high if 0 , low if 1 ).
When the WOC content is higher than SWOF's, an overtaking occurs. The counter value (W0C registers) is automatically reloaded with zero (see Figure 16). If the WOUP bit is high, the new comparison value is reloaded on the shadow SW0R0 registers with the
content of the WORO registers (see Figure 16). This method allows to change frequency and duty cycle without glitch.
Note: If the PWMU0 is off (W0R bit in W0CON not set), W0RnH and W0RnL contents are automatically copied on the shadow registers SWORnH and SWORnLn and the contents of W0FH and W0FL are automatically copied on the shadow registers SW0FH and SW0FL. This allows to charge the correct comparison values for each PWM module as soon as the PWMU0 timer/counter is turned on.

Figure 17. PWMUO Interrupt System


The WOINVn bits that allow output inversion are on the WOMOD (W0 Counter Mode) register (See Table 26.).

Table 31. WORnH: PWMUO module n High Toggle Register WOROH - PWMUO Module 0 High Toggle Register (D9h) W0R1H - PWMU0 Module 1 High Toggle Register (DBh) W0R2H - PWMUO Module 2 High Toggle Register (DDh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W0RnH15 | W0RnH14 | 4 W0RnH13 | W0RnH12 | W0RnH11 | W0RnH10 | WORnH9 | W0RnH8 |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7-0 | $\begin{gathered} \text { W0RnH } \\ 15-8 \end{gathered}$ | PWMUO Module $\mathbf{n}$ high toggle register <br> When the counter exceeds this value, module $n$ output toggles. |  |  |  |  |  |

Reset Value = 0000 0000b
Not bit addressable

Table 32. WORnL: PWMUO module $n$ Low Toggle Register W0ROL - PWMU0 Module 0 Low Toggle Register (DAh)
W0R1L - PWMU0 Module 1 Low Toggle Register (DCh) W0R2H - PWMU0 Module 2 Low Toggle Register (DEh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W0RnL7 | WORnL6 | WORnL5 | W0RnL4 | W0RnL3 | W0RnL2 | W0RnL1 | WORnLO |
| Bit Number | Bit Mnemonic | Description |  |  |  |  |  |
| 7-0 | W0RnL7-0 | PWMUO Module $n$ low toggle register <br> When the counter exceeds this value, module $n$ output toggles. |  |  |  |  |  |

Reset Value $=0000$ 0000b
Not bit addressable

## PWMUO Output Selector

In order to generate no recovery signal, it is possible to configure the microcontroller with the WOOC register to have PWMU0 module 1 XOR PWMU0 module 2 on the W0M1 pin (see Figure 18).

Figure 18. .PWMU0 Output Selector


WOCON and WOMOD are detailed on Table 25 and Table 26.

Each PWMUO module can generate an interrupt. The WOIC register enables or disables interrupt and interrupt flags (See Table 33).

Figure 19. PWMUO Interrupt Configuration


Table 33. PWMUO interrupt control register WOIC - PWMUO Interrupt Control Register (EEh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| WOCF | WOCF2 | WOCF1 | W0CF0 | W0ECOF | WOECF2 | W0ECF1 | W0ECF0 |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | W0COF | PWMUO Counter Overtaking Flag <br> Set by hardware when the counter is higher or equal to SWOF's value. CF flags an interrupt if bit WOECOF is set. WOCOF can be set either by hardware or software but can only be cleared by software. |
| 6 | W0CF2 | PWMUO Module 2 Toggle flag <br> Set by hardware when a match occurs. Can also be set by software. Must be cleared by software. |
| 5 | W0CF1 | PWMUO Module 1 Toggle flag <br> Set by hardware when a match occurs. Can also be set by software. Must be cleared by software. |
| 4 | WOCFO | PWMUO Module 0 Toggle flag <br> Set by hardware when a match occurs. Can also be set by software. Must be cleared by software. |
| 3 | W0ECOF | PWMUO Counter Overtaking flag <br> Set to Enable IT on PWMU0 Counter Overtaking Flag. |
| 2 | W0ECF2 | PWMUO Module 2 Counter flag <br> Set to enable IT on PWMUO Module 2 Toggle flag. |
| 1 | W0ECF1 | PWMUO Module 1 Counter flag <br> Set to enable IT on PWMU0 Module 1Toggle flag. |
| 0 | W0ECFO | PWMUO Module 0 Counter flag <br> Set to enable IT on PWMUO Module OToggle flag. |

Reset Value $=0000$ 0000b
Not bit addressable

PWM Unit 1 (PWMU1)
The PWM unit 1 allows to generate precise pulse width modulation with variable duty cycle and frequency.
The PWMU1 consists of a dedicated 16 bits auto reload counter/timer which serves as a time base for the generation of an independent PWM signal.
Its clock input can be programmed to count any one of the following signals:

- Peripheral clock, CkIdle
- Timer 1 overflow
- External input on W1CI (P4.2)

The PWMU1 timer/counter shares two external I/O. These pins are listed below. If a port is not used for the PWMU1, it can still be used for standard I/O.

| PWMU1 Component | External I/O Pin |
| :---: | :---: |
| 16-bit Counter | W1CI (P4.2) |
| 16 -bit Module 0 | W1M0 (P3.5) |

PWMU1 Timer
The PWMU1 timer is a 16 -bit timer (See Figure 20). The timer count source is determined from the W1CPS1 and W1CPS0 bits in the W1CON register (See Table 34) and can be programmed to run at:

- Peripheral clock, CkIdle
- Timer 1 overflow
- External input on W1CI (P4.2)

The output frequency depends on the timer source and also on the W1F Registers. The timer/counter counts from zero up to a value loaded via SW1F registers. Each time the counter is higher or equal to the SW1F shadow registers value, W1C registers are automatically reloaded with zero. If the W1UP bit is set, the shadow SW1F registers is reloaded with the contents of W1F registers when W1C overtakes. This allows to prevent frequency drift (See Figure 20).
Note: If the PWMU1 is Off (W1R bit in W1CON not set), the contents of W1FH and W1FL are automatically copied on the shadow registers SW1FH and SW1FL. This allows to charge the correct comparison values in order to have the desired frequency as soon as the PWM is turned on.

Figure 20. PWMU1 Timer/Counter


Table 34. W1CON: PWMU1 Control Register W1CON - PWMU1 Control Register (F8h)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W1UP | W1R | - | W1OCLK | W1CPS1 | W1CPS0 | W1INV0 | W1EN0 |


| Bit Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | W1UP | PWMU1 update bit <br> Set by software to request the load of all shadow registers on the next overtaking of the W1C counter. Reset by hardware after the loading of the shadow registers |
| 6 | W1R | PWMU1 Run control bit <br> Set by software to turn the PWMU1 counter on. Must be cleared by software to turn the PWMU1 counter off. |
| 5 | - | Not used |
| 4 | W1OCLK | Output Clocking Control bit. <br> This bit allows to choose between the output clocking signal and the PWM1M0 output. <br> If set, the external clocking is chosen, if clear, PWM1M0 is chosen. |
| 3 | W1CPS1 | PWMU1 Count Pulse Select bit1 |
| 2 | W1CPS0 | PWMU Count Pulse Select bit0 <br> CPS1 CPS0 Selected PWMU1 input <br> 00 Internal clock $\mathrm{f}_{\text {Ckldle }}$ <br> 01 Reserved <br> 10 Timer 1 Overflow <br> 11 External clock input on W 1 Cl at max rate $=\mathrm{f}_{\text {Ckldie }} / 4$ |
| 1 | W1INV0 | PWMU1 Module 0 inverter bit <br> Select the output PWM mode. If set, PWM module 0 output starts with high level. |
| 0 | W1EN0 | PWMU1 Module 0 enable bit <br> Enable PWMU1 module 0 if set. If clear, P3.5 is an I/O port. |

Reset Value = 000'RST_OCLK' 000'RST_OCLK'b Bit addressable

Table 35. W1FH: PWMU1 frequency high control register W1FH - PWMU1 Frequency Control Register (FAh)

| $\mathbf{7}$ | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W1F15 | W1F14 | W1F13 | W1F12 | W1F11 | W1F10 | W1F9 | W1F8 |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| $7-0$ | W1F15-8 | PWMU1 high bits counter control frequency <br> The PWMU1 counter is counting from zero up to W1F15-0 value. |  |  |  |  |  |

Reset Value = 1111 1111b
Not bit addressable

Table 36. W1FL: PWMU1 frequency low control register W1FL - PWMU1 Frequency Control Register (FBh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W1F7 | W1F6 | W1F5 | W1F4 | W1F3 | W1F2 | W1F1 | W1F0 |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7-0 | W1F7-0 | PWMU1 low bits counter control frequency <br> The PWMU1 counter is counting from zero up to W1F15-0 value. |  |  |  |  |  |

Reset Value = 11111111 b
Not bit addressable

Table 37. W1CH: PWMU1 counter high control register W1CH - PWMU1 Counter Control Register (FCh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W1C15 | W1C14 | W1C13 | W1C12 | W1C11 | W1C10 | W1C9 | W1C8 |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7-0 | W1C15-8 | PWMU1 high bits counter frequency |  |  |  |  |  |

Reset Value $=0000$ 0000b
Not bit addressable

Table 38. W1CL: PWMU1 counter low control register W1CL - PWMU1 Counter Control Register (FDh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W1C7 | W1C6 | W1C5 | W1C4 | W1C3 | W1C2 | W1C1 | W1C0 |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| $7-0$ | W1F7-0 | PWMU1 low bits counter frequency |  |  |  |  |  |

Reset Value $=0000$ 0000b
Not bit addressable

All the PWMU1 modules have the same frequency determined by the W1F registers. However, each module has is own duty cycle determined by the W1Rn Registers. ( $n$ is the module number).

When the W1C content is lower than the value programmed via W1Rn registers, the output is the W1INVn-bit (low if 0 , high if 1 ). When it is equal or higher, the output is the opposite of this W1INVn-bit (high if 0 , low if 1 ).
When the W1C content is higher than SW1F's, an overtaking occurs. The counter value (W1C registers) is automatically reloaded with zero (see Figure 21.). If the W1UP bit is high, the new comparison value is reloaded on the shadow SW1R0 registers with the
content of the W1R0 registers (see Figure 21.). This method allows to change frequency and duty cycle without glitch.
Note: If the PWMU1 is Off (W1R bit in W0CON not set), W1RnH and W1RnL contents are automatically copied on the shadow registers SW1RnH and SW1RnLn and the contents of W1FH and W1FL are automatically copied on the shadow registers SW1FH and SW1FL. This allows to charge the correct comparison values for each PWM module as soon as the PWMU1 timer/counter is turned on.

Figure 21. PWMU1 Interrupt System


The W1INV0 bit that allows output inversion is on the W1CON (W1 Control) register (See Table 34.).

Table 39. W1R0H: PWMU1 module 0 High Toggle Register W1R0H - PWMU1 Module 0 High Toggle Register (C9h)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W1R0H15 | W1R0H14 | W1R0H13 | W1R0H12 | W1R0H11 | W1R0H10 | W1R0H9 | W1R0H8 |
| Bit Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7-0 | $\begin{gathered} \text { W1ROH } \\ 15-8 \end{gathered}$ | PWMU1 Module 0 high toggle register <br> When the counter exceeds this value, module 0 output toggles. |  |  |  |  |  |

Reset Value $=0000$ 0000b
Not bit addressable

Table 40. W1R0L: PWMU1 module 0 Low Toggle Register W1R0L - PWMU1 Module 0 Low Toggle Register (CAh)


[^0]Not bit addressable

PWMU1 Output Selector As shown on Figure 22., the PWMU1 can configure P3.5 to be used as

- The PWMU1 module 0 output ( $\mathrm{W} 1 \mathrm{R}=1$ and $\mathrm{W} 1 \mathrm{EN} 0=1$ )
- The External Clocking output $(W 1 O C L K=1$ and $\mathrm{W} 1 E N 0=1)$
- An I/O port (W1EN0 = 0)

This configuration is made via W1CON register (See Table 34.). By default, W1CON register contains 00 h . So P3.5 is configured as an I/O port.
The W1INV0 bit allows to start PWMU1 module 0 period with a high (if set) or low level.
Figure 22. PWMU1 Output Selector


PWMU1 Interrupt System
PWMU1 can generate an interrupt. The W1IC register enables or disables interrupt and interrupt flags (See Table 41).

Figure 23. PWMU1 Interrupt Configuration


Table 41. PWMU1 Interrupt Control Register W1IC - PWMU1 Interrupt Control Register (FEh)

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| W1CF | - | - | W1CF0 | W1ECF | - | - | W1ECF0 |
| Bit <br> Number | Bit Mnemonic | Description |  |  |  |  |  |
| 7 | W1COF | PWMU1 Counter Overtaking Flag <br> Set by hardware when the counter rolls over. CF flags an interrupt if bit W1ECOF is set. W1COF can be set either by hardware or software but can only be cleared by software |  |  |  |  |  |
| 6-5 | - | Not used |  |  |  |  |  |
| 4 | W1CF0 | PWMU1 Module 0 Toggle fla <br> Set by hardware when a match occurs. Can also be set by software. Must be cleared by software. |  |  |  |  |  |
| 3 | W1ECOF | PWMU1 Counter Overtaking Flag <br> Set to Enable PWMU1 Counter Overtaking Flag. |  |  |  |  |  |
| 2-1 | - | Not used |  |  |  |  |  |
| 0 | W1ECF0 | PWMU1 Module 0 Counter flag <br> Set to enable PWMU1 Module OToggle flag. |  |  |  |  |  |

Reset Value $=0000$ 0000b
Not bit addressable

## WatchDog Timer

AT8xEB5114 contains a powerful programmable hardware WatchDog Timer (WDT) that automatically resets the chip if its software fails to reset the WDT before the selected time interval has elapsed. It permits large Time-Out ranking from 16 ms to 2s @Fosc = 12 MHz .

This WDT consists of a 14-bit counter plus a 7-bit programmable counter, a WatchDog Timer reset register (WDTRST) and a WatchDog Timer programmation (WDTPRG) register. When exiting reset, the WDT is -by default- disabled. To enable the WDT, the user has to write the sequence 1EH and E1H into WDRST register. When the WatchDog Timer is enabled, it will increment every machine cycle while the oscillator is running and there is no way to disable the WDT except through reset (either hardware reset or WDT overflow reset). When WDT overflows, it will generate an output RESET pulse at the RST pin. The RESET pulse duration is $96 x T_{\text {OSC }}$, where $T_{O S C}=1 / F_{\text {OSC }}$. To make the best use of the WDT, it should be serviced in those sections of code that will periodically be executed within the time required to prevent a WDT reset.

The WDT is controlled by two registers (WDTRST and WDTPRG)
Figure 24. WatchDog Timer


Figure 25. WDTPRG Register
WDTPRG - WatchDog Timer Duration Programming register (A7h).

| $\begin{gathered} \text { Bit } \\ \text { Number } \end{gathered}$ | Bit <br> Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 6 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 5 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 4 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 3 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 2 | S2 | WatchDog Timer Duration selection bit 2 Work in conjunction with bit 1 and bit 0 . |
| 1 | S1 | WatchDog Timer Duration selection bit 1 Work in conjunction with bit 2 and bit 0 . |
| 0 | SO | WatchDog Timer Duration selection bit 0 Work in conjunction with bit 1 and bit 2. |

Reset Value $=$ XXXX X000b

The three lower bits (S0, S1, S2) located into WDTPRG register permit to program the WDT duration.

| S2 | S1 | S0 | Machine Cycle Count |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | $2^{14}-1$ |
| 0 | 0 | 1 | $2^{15}-1$ |
| 0 | 1 | 0 | $2^{16}-1$ |
| 0 | 1 | 1 | $2^{17}-1$ |
| 1 | 0 | 0 | $2^{18}-1$ |
| 1 | 0 | 1 | $2^{19}-1$ |
| 1 | 1 | 0 | $2^{20}-1$ |
| 1 | 1 | 1 | $2^{21}-1$ |

To compute WD Time-Out, the following formula is applied:

$$
\text { TimeOut }=\frac{\left((\text { FclkPeriph })^{x 2}\right)^{\times 2}}{12 \times\left(\left(2^{14} \times 2^{\text {Svalue }}\right)-1\right) \times(15-\text { CKRL })}
$$

Note: Note: Value represents the decimal value of (S2 S1 S0) / CKRL represents the Prescaler

Find Hereafter computed Time-Out value for Fosc $=12 \mathrm{MHz}$

Table 42. Time-Out Computation @ 12 MHz

| $\mathbf{S 2}$ | $\mathbf{S 1}$ | $\mathbf{S 0}$ | Time-Out for $\mathbf{F}_{\text {osc }}=\mathbf{1 2} \mathbf{~ M H z}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 16.38 ms |
| 0 | 0 | 1 | 32.77 ms |
| 0 | 1 | 0 | 65.54 ms |
| 0 | 1 | 1 | 131.07 ms |
| 1 | 0 | 0 | 262.14 ms |
| 1 | 0 | 1 | 524.29 ms |
| 1 | 1 | 0 | 1.05 s |
| 1 | 1 | 1 | 2.10 s |

Table 43. WDTRST Register

| 7 | 6 | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - | - | - | - | - | - |

Reset Value $=$ XXXX XXXXb

The WDTRST register is used to reset/enable the WDT by writing 1 EH then E 1 H in sequence.

## WatchDog Timer During Power Down Mode and Idle

In Power Down mode the oscillator stops, which means the WDT also stops. While in Power Down mode the user does not need to service the WDT. There are 2 methods of exiting Power Down mode: by a hardware reset or via a level activated external interrupt which is enabled prior to entering Power Down mode. When Power Down is exited with hardware reset, servicing the WDT should occur as it normally does whenever AT8xEB5114 is reset. Exiting Power Down with an interrupt is significantly different. The interrupt is held low long enough for the oscillator to stabilize. When the interrupt is brought high, the interrupt is serviced. To prevent the WDT from resetting the device while the interrupt pin is held low, the WDT is not started until the interrupt is pulled high. It is suggested that the WDT be reset during the interrupt service for the interrupt used to exit Power Down.

To ensure that the WDT does not overflow within a few states of exiting of power down, it is best to reset the WDT just before entering power down.
In the Idle mode, the oscillator continues to run. To prevent the WDT from resetting while in Idle mode, the user should always set up a timer that will periodically exit Idle, service the WDT, and re-enter Idle mode.

## Analog-to-Digital Converter (ADC)

## Features

This section describes the on-chip 10 bit analog-to-digital converter of the AT8xEB5114. Six ADC channels are available for sampling of the external sources AIN0 to AIN5. An analog multiplexer allows the single ADC converter to select one from the 6 ADC channels as ADC input voltage (ADCIN). ADCIN is converted by the 10 bit-cascaded potentiometric ADC.
8 to 10 bits resolution can only be reached while using an external voltage reference.
For the precision conversion, set bits PSIDLE and ADSST in ADCON register to start the conversion. The chip is in a pseudo-idle mode, the CPU doesn't run but the peripherals are always running. This mode allows digital noise to be lower, to ensure precise conversion.

For accurate conversion, set bits QUIETM and ADSST in ADCON register to start the conversion. The chip is in a quiet mode, the AD is the only peripheral running. This mode allows digital noise to be as low as possible, to ensure high precision conversion.

For these modes it is necessary to work with end of conversion interrupt, which is the only way to wake up the chip.
If another interrupt occurs during the precision conversion, it will be treated only after this conversion is ended.

- 6 channels with multiplexed inputs
- One channel with input signal average extraction and programmable gain amplification.
- 10-bit cascaded potentiometric ADC
- Typical conversion time 20 micro-seconds
- Zero Error (offset) +/- 2 LSB max
- External Positive Reference Voltage Range 2.4 to Vcc
- Internal Positive Reference typical Voltage 2.4 Volt ${ }^{(1)}$
- ADCIN Range 0 to $\mathrm{V}_{\text {REF }}$
- Integral non-linearity typical 1 LSB ${ }^{(1)}$
- Differential non-linearity typical 0.5 LSB ${ }^{(1)}$
- Conversion Complete Flag or Conversion Complete Interrupt
- Selected ADC Clock

Note: (1): See "DC Parameters for A/D Converter" on page 88.

## ADC I/O Functions

AINx are general I/O that are shared with the ADC channels. The channel select bit in ADCF register define which ADC channel pin will be used as ADCIN. The remaining ADC channels pins can be used as general purpose I/O or as the alternate function that is available. Writings to the port registers which aren't selected by the ADCF will not have any effect.

Figure 26. ADC Description


Figure 27 shows the timing diagram of a complete conversion. For simplicity, the figure depicts the waveforms in idealized form and do not provide precise timing information. For ADC characteristics and timing parameters refer to theSection "DC Parameters for A/D Converter", page 88.

Figure 27. Timing Diagram


Note: $\quad$ Tsetup $=0$ CLK

## Channel 3 Amplifier and Rectifying Function

If needed, the average value of the rectified signal on channel 3 can be extracted and amplified before A/D conversion as shown on Figure 28.

Figure 28. Channel 3 Amplifier


The main characteristics of this block are the following:

- Input signal level: sine wave centered around Vssa, peak value from 70 to 550 mV depending on gain, Frequency range from 35 to 70 KHz . Be sure that the peak value on the amplifier output is lower than voltage supply.
- Gain: x5, x10, x15 and x20, selected using AC3E, AC3G1 and AC3G0 in ADC Amplifier register (See Table 44 and Table 52)
- Max time constant of the average value extraction: 0.5 ms . When the gain is changed or when the signal levels changes from the minimum to the maximum value, a new measurement can be done after 10 time constant.
- The amplifier needs 20us to fully load the ADC hold capacitance so the ADC conversion must occurs at least 20us after the amplified channel is sampled.
- Accuracy on amplification and extraction: +/- 5\%

Note: The AIN3 direct channel is not equivalent to the other channels. There is a serial resistance of around $100 \mathrm{~K} \Omega$ between the pin and the ADC input. So when the amplifier is bypassed, it is necessary to switch at least 20us the mux on AIN3 input before starting a conversion.
Table 44. ADCF Register


Reset Value $=0000$ 0000b

ADC Converter Operation

A start of single A/D conversion is triggered by setting bit ADSST (ADCON.3).
The busy flag ADSST(ADCON.3) remains set as long as an $A / D$ conversion is running. After completion of the A/D conversion, it is cleared by hardware. When a conversion is running, this flag can be read only, a write has no effect.

The end-of-conversion flag ADEOC (ADCON.4) is set when the value of conversion is available in ADDH and ADDL, it is cleared by software. If the bit EADC (IE0.6) is set, an interrupt occur when flag ADEOC is set (see Figure 30). Clear this flag for re-arming the interrupt.

The bits SCH0 to SCH2 in ADCON register are used for the analog input channel selection.

Before starting normal power reduction modes the ADC conversion has to be completed.

Table 45. Selected Analog Input

| SCH2 | SCH1 | SCH0 | Selected Analog input |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | AN0 |
| 0 | 0 | 1 | AN1 |
| 0 | 1 | 0 | AN2 |
| 0 | 1 | 1 | AN3 |
| 1 | 0 | 0 | AN4 |
| 1 | 0 | 1 | AN5 |
| 1 | 1 | 0 |  |
| 1 | 1 | 1 |  |

## Voltage Conversion

## Clock Selection

When the ADCIN is equal to VAREF the ADC converts the signal to 3FFh (full scale). If the input voltage equals VAGND, the ADC converts it to 000h. Input voltage between VAREF and VAGND are a straight-line linear conversion. All other voltages will result in 3FFh if greater than VAREF and 000h if less than VAGND.
Note that ADCIN should not exceed VAREF absolute maximum range.
The maximum clock frequency for ADC (CONV_CK for Conversion Clock) is defined in the Section "AC Parameters", page 88. A prescaler is featured to generate the CONV_CK clock from the oscillator frequency. The prescaler value PRS[6:0] is defined in the ADCLK register (see Table 49 on page 64)

Figure 29. A/D Converter clock


The conversion frequency CONV_CK is derived from the oscillator frequency with the following formulas:

$$
\begin{aligned}
\mathrm{F}_{\text {CKAdc }} & =\mathrm{F}_{\text {OscOut }} /\left(32-2^{*} \mathrm{CKRL}\right) \text {, if } \mathrm{X} 2=0 \\
& =\mathrm{F}_{\text {Oscout }} \quad \text {, if } \mathrm{X} 2=1
\end{aligned}
$$

and
$\mathrm{F}_{\text {CONV_CK }}=\mathrm{F}_{\text {CKAdc }} /\left(2^{*} \mathrm{PRS}\right)$, if $\mathrm{PRS}>0$
$\mathrm{F}_{\text {CONV_CK }}=\mathrm{F}_{\text {CKAdc }} / 256$, if $\mathrm{PRS}=0$
Some examples can be found on table below:

| $\mathbf{F}_{\text {OscOut }}$ <br> $\mathbf{M H z}$ | X2 | CKRL | $\mathbf{F}_{\text {CKAdc }}$ <br> $\mathbf{M h z}$ | PRSw | $\mathbf{F}_{\text {CoNv_ck }}$ <br> $\mathbf{k h z}$ | Conversion <br> time $\boldsymbol{\mu s}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 16 | 0 | F | 8 | 12 | 333 | 33 |
| 16 | 1 | NA | 16 | 32 | 250 | 44 |
|  |  |  |  |  |  |  |

ADC Standby Mode

Voltage Reference

## IT ADC Management

When the ADC is not used, it is possible to set it in standby mode by clearing bit ADEN in ADCON register.

The voltage reference can be either internal or external.
As input, the $\mathrm{V}_{\text {REF }}$ pin is used to enter the voltage reference for the $\mathrm{A} / \mathrm{D}$ conversion.
When the voltage reference is active, the $\mathrm{V}_{\text {REF }}$ pin is an output. This voltage can be used for the A/D and for any other application requiring a voltage independent from the power supply. Voltage typical value is 2.4 volt (See "DC Parameters for A/D Converter" on page 88.)

An interrupt end-of-conversion will occur when the bit ADEOC is activated and the bit EADC is set. For re-arming the interrupt the bit ADEOC must be cleared by software.

Figure 30. ADC Interrupt Structure


## Accuracy improvement on analog to digital conversion using the internal voltage reference

## Overview

Coefficient address

## Coefficient format

The internal Vref absolute accuracy is around 4\%. This variation is mainly due to the temperature, the process, and the voltage variations. In order to increase the accuracy of the measurements made thanks to the ADC, it is possible to make a software correction of the Vref, in order to calculate the result the ADC should have returned in case the Vref was more accurate.

The idea of this improvement is the following: Because there is an EEPROM stacked on the product, it is possible to store a linear coefficient which allow a correction of the process variations.

The coefficient is stored at the address $0 \times 00$ of the serial data EEPROM stacked on the AT8xEB5114.

In order to ease the calculation, this coefficient has been stored as a floating decimal number corresponding to Table 46.

Table 46. Calibration coefficient storage format

| Bit | Value |
| :---: | :---: |
| 7 | 1, |
| 6 | $1 / 2$ |
| 5 | $1 / 4$ |
| 4 | $1 / 8$ |
| 3 | $1 / 16$ |
| 2 | $1 / 32$ |
| 1 | $1 / 64$ |
| 0 | $1 / 128$ |

It means that if the value is $0 \times 80$, the coefficient is equal to 1 . If the coefficient is $0 \times 7 e$, the coefficient is equal to 0,1111110 in binary which is 0,983 in decimal.

During the test, the Vref is measured, and the calibration value calculated is stored at the address $0 \times 00$ of the stack die in accordance with the Table 46 format value.

The relation between the coefficient stored, and the true Vref measurement are recorded on the Table 47.

Table 47. Relation between True Vref measurement and coefficient stored into the EEPROM

| True Vref <br> Value (V) Min | 2.300 | 2.316 | 2.334 | 2.353 | 2.372 | 2.391 | 2.409 | 2.428 | 2.447 | 2.466 | 2.484 |  |
| :--- | :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Typ | 2.306 | 2.325 | 2.345 | 2.362 | 2.381 | 2.400 | 2.419 | 2.438 | 2.456 | 2.475 | 2.494 |
|  | Max | 2.316 | 2.334 | 2.353 | 2.372 | 2.391 | 2.409 | 2.428 | 2.447 | 2.466 | 2.484 | 2.500 |
| Value stored |  | $0 \times 7 \mathrm{~b}$ | $0 \times 7 \mathrm{c}$ | $0 \times 7 \mathrm{~d}$ | $0 \times 7 \mathrm{e}$ | $0 \times 7 \mathrm{f}$ | $0 \times 80$ | $0 \times 81$ | $0 \times 82$ | $0 \times 83$ | $0 \times 84$ | $0 \times 85$ |
| decimal value |  | 0.961 | 0.969 | 0.977 | 0.984 | 0.992 | 1 | 1.008 | 1.016 | 1.023 | 1.031 | 1.039 |

## How to Take Advantage of the Calibration Value

The coefficient stored on the stacked die allow to determine the conversion result the AT8xEB5114 should have returned in case its Vref was exactly equal to 2.4 V . In order to determine it, a multiplication of the result of the conversion with the coefficient stored in the stack, followed by a shift are sufficient.

## Example

## Assembler code example

## Registers

Vref is 2.36 V instead of 2.4 V , and only 8 bits are necessary.
The value measured during the test 2.36 V . So, in accordance with the Table 47, the coefficient which has to be stored on the EEPROM is $0 x 7 \mathrm{e}$ which corresponds to 0.1111110 in binary, which also corresponds to around 2.36/2.4.

If, for example, after a conversion, the ADDH register contains 0xf0, to know the result the ADC should have returned in case the Vref was really at 2.4 V , the following operations are necessary:
$0 x f 0 * 0 x 7 e=11110000 * 01111110=0 x 7620=0111011000100000$.
So because of the point on the coefficient, the result is 1110110 which is $0 x e c$.
This is an example of assembler code optimized for size and fast recalculation in case 8 bits are sufficient.

| start_adjustement : | MOV B, Coeff ; Coeff |  |
| :--- | :--- | :--- |
|  | MOV A, ADDH ; ADC result |  |
|  | MUL AB | $;$ |
|  | RLC A | ; Recover lowest bit |
|  | MOV A, B | ; |
|  | RLC A | ; Recover result |
|  | JNC end_fix ; Result OK |  |
| end_adjustement | MOV A, \#Offh ; Overflow |  |
|  | RET |  |

The new result is stored on the accumulator.
This routine requires 15 bytes +3 bytes for the long call (LCALL).
The execution of the subroutine (including the LCALL) is 18 cycles in normal case and 19 cycles in case of overflow (less than 10us with a 12 MHz oscillator and the X2 mode).

Table 48. ADCON Register
ADCON (S:F3h)
ADC Control Register

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| QUIETM | PSIDLE | ADEN | ADEOC | ADSST | SCH2 | SCH1 | SCH0 |
| Bit Number | Bit Mnemonic | Description |  |  |  |  |  |
| 7 | QUIETM | Quiet mode (best precision) <br> Set to put in quiet mode during conversion. <br> Cleared by hardware after completion of the conversion. |  |  |  |  |  |
| 6 | PSIDLE | Pseudo Idle mode (good precision) <br> Set to put in idle mode during conversion. <br> Cleared by hardware after completion of the conversion. |  |  |  |  |  |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :--- |
| 5 | ADEN | Enable/Standby Mode <br> Set to enable ADC. <br> Clear for Standby mode. |
| 4 | ADEOC | End Of Conversion <br> Set by hardware when ADC result is ready to be read. This flag can generate an <br> interrupt. <br> Must be cleared by software. |
| 3 | ADSST | Start and Status <br> Set to start an A/D conversion. <br> Cleared by hardware after completion of the conversion. |
| $2-0$ | SCH2:0 | Selection of channel to convert <br> see Table 45 on page 60. |

Reset Value $=$ X000 0000b

Table 49. ADCLK Register
ADCLK (S:F2h)
ADC Clock PrescalersC

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| SELREF | PRS 6 | PRS 5 | PRS 4 | PRS 3 | PRS 2 | PRS 1 | PRS 0 |
| $\begin{gathered} \text { Bit } \\ \text { Number } \end{gathered}$ | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7 | SELREF | Selection and activation of the internal 2.4 V voltage reference Set to enable the internal voltage reference. Clear to disable the internal voltage reference. |  |  |  |  |  |
| 6-0 | PRS6:0 | Clock Prescaler$\begin{aligned} & \mathrm{f}_{\mathrm{CONV} \text { _CK }}=\mathbf{f}_{\mathrm{CKADC}} /(2 * \text { PRS }) \\ & \text { if PRS }=0, \mathrm{f}_{\mathrm{CONV} \text { _CK }}=\mathbf{f}_{\mathrm{CKADC}} / 256 \end{aligned}$ |  |  |  |  |  |

Reset Value $=0000$ 0000b

Table 50. ADDH Register
ADDH (S:F5h Read Only)
ADC Data High byte register

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ADAT 9 | ADAT 8 | ADAT 7 | ADAT 6 | ADAT 5 | ADAT 4 | ADAT 3 | ADAT 2 |
| Bit Number | Bit Mnemonic | Description |  |  |  |  |  |
| 7-0 | ADAT9:2 | ADC result bits 9-2 |  |  |  |  |  |

Read only register
Reset Value $=00 \mathrm{~h}$

Table 51. ADDL Register ADDL (S:F4h Read Only) ADC Data Low byte register

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | - |  | - | - | ADAT 1 | ADAT 0 |
| Bit Number | Bit <br> Mnemonic | Description |  |  |  |  |  |
| 7-6 | - | Reserved <br> The value read from these bits are indeterminate. Do not set these bits. |  |  |  |  |  |
| 1-0 | ADAT1:0 | ADC result bits 1-0 |  |  |  |  |  |

Read only register
Reset Value = xxxx xx00b

Table 52. ADCF Register
ADCF (S:F6h)
ADC Configuration

| 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | - | CH 5 | CH 4 | CH 3 | CH 2 | CH 1 | CH 0 |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :--- |
| $7-6$ | - | Not used |
| 5 | CH5 | Channel Configuration <br> Set to use P3.4 as ADC input <br> Clear to use P3.4 as an other function |
| 4 | CH4 | Channel Configuration <br> Set to use P3.3 as ADC input <br> Clear to use P3.3 as an other function |
| $3-0$ | CH3-0 | Channel Configuration <br> Set to use P4.x as ADC input <br> Clear to use P4.x as an other function |

Reset Value $=0000$ 0000b

## Interrupt System

The AT8xEB5114 has a total of 8 interrupt vectors: two external interrupts ( $\overline{\text { INT0 }}$ and INT1), two timer interrupts (timers 0, 1), serial port interrupt, PWMU0, PWMU1 and A/D. These interrupts are shown in Figure 31.

Figure 31. Interrupt Control System


Each of the interrupt sources can be individually enabled or disabled by setting or clearing a bit in the Interrupt Enable register (See Table 54). This register also contains a global disable bit, which must be cleared to disable all interrupts simultaneously.
Each interrupt source can also be individually programmed to one of four priority levels by setting or clearing a bit in the Interrupt Priority register (See Table 55) and in the Interrupt Priority High register (See Table 56). Table 53 shows the bit values and priority levels associated with each combination.

Table 53. Priority Bit Level Values

| IPH.x | IP.x | Interrupt Level Priority |
| :---: | :---: | :---: |
| 0 | 0 | 0 (Lowest) |
| 0 | 1 | 1 |
| 1 | 0 | 2 |
| 1 | 1 | 3 (Highest) |

A low-priority interrupt can be interrupted by a high priority interrupt, but not by another low-priority interrupt. A high-priority interrupt can't be interrupted by any other interrupt source.
If two interrupt requests of different priority levels are received simultaneously, the request of higher priority level is serviced. If interrupt requests of the same priority level are received simultaneously, an internal polling sequence determines which request is serviced. Thus within each priority level there is a second priority structure determined by the polling sequence.

| Interrupt Name | Interrupt Address Vector | Priority Number |
| :---: | :---: | :---: |
| external interrupt (INT0) | 0003 h | 1 |
| Timer0 (TF0) | 000 Bh | 2 |
| external interrupt (INT1) | 0013 h | 3 |
| Timer1 (TF1) | 001 Bh | 4 |
| PWM0 | 0023 h | 5 |
| PWM1 | 002 Bh | 6 |
| ADC | 0033 h | 7 |

Table 54. IENO Register
IENO - Interrupt Enable Register (A8h)

| $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| EA | EADC | EW1 | EW0 | ET1 | EX1 | ET0 | EX0 |


| Bit Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | EA | Enable All interrupt bit <br> Clear to disable all interrupts. <br> Set to enable all interrupts. <br> If $E A=1$, each interrupt source is individually enabled or disabled <br> by setting or clearing its interrupt enable bit. |
| 6 | EADC | ADC Interrupt Enable Clear to disable the ADC interrupt. Set to enable the ADC interrupt. |
| 5 | EW1 | PWM1 Enable bit Clear to disable PWMU interrupt. Set to enable PWMU port interrupt. |
| 4 | EW0 | PWMO Enable bit Clear to disable PWMU interrupt. Set to enable PWMU port interrupt. |
| 3 | ET1 | Timer 1 overflow interrupt Enable bit Clear to disable timer 1 overflow interrupt. Set to enable timer 1 overflow interrupt. |
| 2 | EX1 | External interrupt 1 Enable bit Clear to disable external interrupt 1. Set to enable external interrupt 1. |
| 1 | ETO | Timer 0 overflow interrupt Enable bit Clear to disable timer 0 overflow interrupt. Set to enable timer 0 overflow interrupt. |
| 0 | EXO | External interrupt 0 Enable bit Clear to disable external interrupt 0 . Set to enable external interrupt 0 . |

Reset Value $=0000$ 0000b
Bit addressable

Table 55. IPL0 Register
IPLO - Interrupt Priority Register (B8h)

| $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | PADC | PW1 | PW0 | PT1 | PX1 | PT0 | PX0 |


| Bit <br> Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :--- |
| 7 | - | ReservedThe value read from this bit is indeterminate. Do not set this bit. |
| 6 | PADC | ADC interrupt Priority bit <br> Refer to PADCH for priority level |
| 5 | PW1 | PWMU1 Priority bit <br> Refer to PW1H for priority level. |
| 4 | PW1 | PWMU0 Priority bit <br> Refer to PW1H for priority level. |
| 3 | PT1 | Timer 1 overflow interrupt Priority bit <br> Refer to PT1H for priority level. |
| 2 | PT0 | External interrupt 1 Priority bit <br> Refer to PX1H for priority level. |
| 1 | PX0Rerflow interrupt Priority bit <br> Refer to PTOH for priority level. |  |
| 0 | External interrupt 0 Priority bit <br> Refer to PX0H for priority level. |  |

Reset Value = X000 0000b
Bit addressable.

Table 56. IPHO Register
IPH0 - Interrupt Priority High Register (B7h)

| $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| - | PADCH | PW1H | PWOH | PT1H | PX1H | PTOH | PX0H |


| Bit Number | Bit <br> Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | - | Reserved <br> The value read from this bit is indeterminate. Do not set this bit. |
| 6 | PADCH | ADC Interrupt Priority level most significant bit |
| 5 | PW1H | PWMU1 Priority High bit |
| 4 | PW1H | PWMUO Priority High bit |
| 3 | PT1H | Timer 1 overflow interrupt Priority High bit |
| 2 | PX1H | External interrupt 1 Priority High bit |
| 1 | PTOH | Timer 0 overflow interrupt Priority High bit |
| 0 | PXOH | External interrupt 0 Priority High bit |

Reset Value $=$ X000 0000b
Not bit addressable

## Flash Memory

## FMO Memory

Architecture

## User Space

Extra Row (XRow)

Hardware security Byte

## Column latches

## Overview of Flash Memory Operations

As shown Figure 32, the Flash version of AT8xEB5114 implements 4 Kbytes of on-chip program/code memory.
The Flash memory increases EPROM and ROM functionality by in-circuit electrical erasure and programming. Thanks to the internal charge pump, the high voltage needed for programming or erasing Flash cells is generated on-chip using the standard VDD voltage.
Hardware programming mode is available using specific programming tool.
AT8xEB5114 features a Flash memory containing 4 Kbytes of program memory (user space) organized into 128 byte pages,

This Flash memory is programmable by parallel programming.
Figure 32. Flash Memory Architecture


The Flash memory is made up of 4 blocks (see Figure 32):

- The memory array (user space) 4 Kbytes
- The Extra Row
- The Hardware security bits
- The column latch registers

This space is composed of a 4 Kbytes Flash memory organized in 32 pages of 128 bytes. It contains the user's application code.

This row is a part of flash memory and has a size of 128 bytes. The extra row may contain information for boot loader usage.

The Hardware Security Byte space is a part of flash memory and has a size of 1 byte. The 4 MSB can be read/written by software, the 4 LSB can only be read by software and written by hardware in parallel mode.

The column latches, also part of flash memory, have a size of full page (128 bytes). The column latches are the entrance buffers of the three previous memory locations (user array, XROW and Hardware security byte).

The CPU interfaces to the Flash memory through the FCON register used to:

- Map the memory spaces in the adressable space
- Launch the programming of the memory spaces
- Get the status of the flash memory (busy/not busy)

Mapping of the Memory Space

By default, the user space is accessed by MOVC instruction for read only. The column latches space is made accessible by setting the FPS bit in FCON register. Writing is possible from 0000h to 0FFFh, address bits 6 to 0 are used to select an address within a page while bits 14 to 7 are used to select the programming address of the page.

The other memory spaces (user, extra row, hardware security) are made accessible in the code segment by programming bits FMOD0 and FMOD1 in FCON register in accordance with Table 57. A MOVC instruction is then used for reading these spaces.

Table 57. .FM0 Blocks Select Bits

| FMOD1 | FMOD0 | FM0 Adressable space |
| :---: | :---: | :---: |
| 0 | 0 | User (0000h-FFFFh) |
| 0 | 1 | Extra Row(FF80h-FFFFh) |
| 1 | 0 | Hardware Security Byte (0000h) |
| 1 | 1 | reserved |

Launching programming
FPL3:0 bits in FCON register are used to secure the launch of programming. A specific sequence must be written in these bits to unlock the write protection and to launch the programming. This sequence is 5xh followed by Axh. Table 33 summarizes the memory spaces to program according to FMOD1:0 bits.

Figure 33. Programming spaces

|  | Write to FCON |  |  |  | Operation |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | FPL3:0 | FPS | FMOD1 | FMODO |  |
| User | 5 | X | 0 | 0 | No action |
|  | A | X | 0 | 0 | Write the column latches in user space |
| Extra Row | 5 | X | 0 | 1 | No action |
|  | A | X | 0 | 1 | Write the column latches in extra row space |
| Hardware Security Byte | 5 | X | 1 | 0 | No action |
|  | A | X | 1 | 0 | Write the fuse bits space |
| Reserved | 5 | X | 1 | 1 | No action |
|  | A | X | 1 | 1 | No action |

Notes: 1. The sequence $5 x h$ and Axh must be executing without instructions between then otherwise the programming is aborted.
2. Interrupts that may occur during programming time must be disable to avoid any spurious exit of the idle mode.

Status of the Flash Memory

Loading the Column Latches

The bit FBUSY in FCON register is used to indicate the status of programming. FBUSY is set when programming is in progress.

Any number of data from 1 byte to 128 bytes can be loaded in the column latches. This provides the capability to program the whole memory by byte, by page or by any number of bytes in a page.
When programming is launched, an automatic erase of the locations loaded in the column latches is first performed, then programming is effectively done. Thus no page or block erase is needed and only the loaded data are programmed in the corresponding page.
The following procedure is used to load the column latches and is summarized in Figure 34:

- Disable interrupt and map the column latch space by setting FPS bit.
- Load the DPTR with the address to load.
- Load Accumulator register with the data to load.
- Execute the MOVX @DPTR, A instruction.
- If needed loop the three last instructions until the page is completely loaded.
- unmap the column latch and Enable Interrupt

Figure 34. Column Latches Loading Procedure


Note: The last page address used when loading the column latch is the one used to select the page programming address.

User
The following procedure is used to program the User space and is summarized in Figure 35:

- Load data in the column latches from address 0000 h to $0 F F F h^{1}$.
- Disable the interrupts.
- Launch the programming by writing the data sequence 50 h followed by AOh in FCON register (only from FM1).
The end of the programming indicated by the FBUSY flag cleared.
- Enable the interrupts.

Note: 1. The last page address used when loading the column latch is the one used to select the page programming address.

## Extra Row

The following procedure is used to program the Extra Row space and is summarized in Figure 35:

- Load data in the column latches from address FF80h to FFFFh.
- Disable the interrupts.
- Launch the programming by writing the data sequence 52 h followed by A2h in FCON register (only from FM1).
The end of the programming indicated by the FBUSY flag cleared.
- Enable the interrupts.

Figure 35. Flash and Extra row Programming Procedure


Hardware Security Byte
The following procedure is used to program the Hardware Security Byte space and is summarized in Figure 36:

- Set FPS and map Hardware byte ( $\mathrm{FCON}=0 \times 0 \mathrm{C}$ )
- Disable the interrupts.
- Load DPTR at address 0000h.
- Load Accumulator register with the data to load.
- Execute the MOVX @DPTR, A instruction.
- Launch the programming by writing the data sequence 54h followed by A4h in FCON register (only from FM1).
The end of the programming indicated by the FBusy flag cleared.
- Enable the interrupts.

Figure 36. Hardware Programming Procedure


User
The following procedure is used to read the User space and is summarized in Figure 37:

- Map the User space by writing 00h in FCON register.
- Read one byte in Accumulator by executing MOVC A,@A+DPTR with A=0 \& DPTR= 0000h to FFFFh.


## Extra Row

The following procedure is used to read the Extra Row space and is summarized in Figure 37:

- Map the Extra Row space by writing 02h in FCON register.
- Read one byte in Accumulator by executing MOVC A,@A+DPTR with A=0 \& DPTR= FF80h to FFFFh.
- Clear FCON to unmap the Extra Row.

Hardware Security Byte

The following procedure is used to read the Hardware Security space and is summarized in Figure 37:

- Map the Hardware Security space by writing 04h in FCON register.
- Read the byte in Accumulator by executing MOVC A,@A+DPTR with A= 0 \& DPTR= 0000h.
- Clear FCON to unmap the Hardware Security Byte.

Figure 37. Reading Procedure


Flash Protection from Parallel Programming

The three lock bits in Hardware Security Byte are programmed according to Table 58, will provide different level of protection for the on-chip code and data located in flash memory.
The only way for write this bits are the parallel mode.
Table 58. Program Lock Bit

| Program Lock Bits |  | Protection Description |  |
| :---: | :---: | :---: | :--- |
| Security <br> level | LB1 | LB0 |  |
| 1 | U | U | No program lock feature enabled. |
| 2 | U | P | Writing Flash data from programmer is disabled but still allowed from <br> internal code execution. |
| 3 | P | U | Writing and reading Flash data from programmer is disabled but still <br> allowed from internal code execution. |

WARNING: Security level 2 and 3 should only be programmed after Flash and Core verification.

Table 59. FCON: Flash Control Register
FCON - Flash Control Register (D1h)

| 7 | 5 | 4 | 3 | 2 | 1 |  |
| :---: | :---: | :--- | :---: | :---: | :---: | :---: |
| FPL3 | FPL2 | FPL1 | FPL0 | FPS | FMOD1 | FMOD0 |
| Bit <br> Number | Bit <br> Mnemonic | Description |  |  |  |  |
| $7-4$ | FPL3:0 | Programming Launch Command Bits <br> Write 5Xh followed by AXh to launch the programming according to FMOD1:0. <br> (see Figure 33.) |  |  |  |  |
| 3 | FPS | Flash Map Program Space <br> Set to map the column latch space in the data memory space. <br> Clear to re-map the data memory space. |  |  |  |  |
| $2-1$ | FMOD1:0 | Flash Mode <br> See Table 57 or Table 33. |  |  |  |  |
| 0 | FBUSY | Flash Busy <br> Set by hardware when programming is in progress. <br> Clear by hardware when programming is done. <br> Can not be cleared by software. |  |  |  |  |

Reset Value $=0000$ 0000b

## AT8xEB5114 ROM

## ROM Structure

The AT8xEB5114 ROM memory is divided in two different arrays:

- the code array: 4 Kbytes.
- the configuration byte:1 byte.

Hardware Configuration Byte
The configuration byte sets the starting microcontroller options and the security levels. The starting default options are X1 mode, Oscillator A.

Table 60. Hardware Security Byte (HSB)
HSB (S:EFh)
Power configuration Register

| 7 | $\mathbf{7}$ | $\mathbf{6}$ | $\mathbf{5}$ | $\mathbf{4}$ | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| X2 | RST_OSC1 | RST_OSC0 | RST_OCLK | - | - | LB1 | LB0 |


| Bit Number | Bit Mnemonic | Description |
| :---: | :---: | :---: |
| 7 | X2 | X2 Mode <br> Clear to force X2 mode (CkOut = OscOut) <br> Set to use the prescaler mode (CkOut = OscOut / (2*(16-M))) |
| 6 | RST_OSC1 | Oscillator bit 1 on reset |
| 5 | RST_OSC0 | Oscillator bit 0 on reset <br> Oscillator bit on reset <br> 11: allow OSCA <br> 10: allow OSCB <br> 01: allow OSCC <br> 00: reserved |
| 4 | RST_OCLK | Output clocking signal after RESET <br> Clear to start the microcontroller with a low level on P3.5 followed by an output clocking signal on P3.5 as soon as the microcontroller is started. This signal has is a $1 / 3$ high $2 / 3$ low signal. Its frequency is equal to (CKout / 3 ). Set to start on normal conditions: No signal on P3.5 which is pulled up. |
| 3 | CKRLRV | CKRL Reset Value <br> If set, the microcontroller starts with the prescaler reset value $=$ XXXX 1000 (OscOut = CkOut/16). <br> If clear, the microcontroller starts with a prescaler reset value $=$ XXXX 1111 (OscOut = CkOut/2). |
| 2 | - | Reserved |
| 1-0 | LB1-0 | User Program Lock Bits See Table 61 on page 81 |

HSB = 1111 XX11b
Note: Whatever the value of RST_OSC, the XTAL1 input is always validated in order to enter in test modes.

ROM Lock System

Program ROM lock Bits

The program Lock system, when programmed, protects the on-chip program against software piracy.

The lock bits when programmed according to Table 61 will provide different level of protection for the on-chip code and data.
Table 61. Program Lock bits

| Program Lock Bits |  | Protection Description |  |
| :---: | :---: | :---: | :--- |
| Security <br> level | LB1 | LB0 |  |
| 1 | U | U | No program lock feature enabled. |
| 3 | P | U | Reading ROM data from programmer is disabled. |

U: unprogrammed
P: programmed

## Stacked EEPROM

## Overview

## Protocol

The AT8xEB5114 features a stacked 2-wire serial data EEPROM. The data EEPROM allows to save up to 256 bytes. The EEPROM is internally connected to P3.6 and P3.7 which are respectively connected to the SDA and the SCL pins.

In order to access this memory, it is necessary to use software subroutines according to the AT24C02 datasheet. Nevertheless, because the internal pull-up resistors of the AT8xEB5114 is quite high (around $100 \mathrm{~K} \Omega$ ), the protocol should be slowed in order to be sure that the SDA pin can rise to the high level before reading it.
Another solution to keep the access to the EEPROM in specification is to work with a software pull-up.
Using a software pull-up, consists of forcing a low level at the output pin of the microcontroller before configuring it as an input (high level).
The C51 the ports are "quasi-bidirectional" ports. It means that the ports can be configured as output low or as input high. In case a port is configured as an output low, it can sink a current and all internal pull-ups are disconnected. In case a port is configured as an input high, it is pulled up with a strong pull-up (a few hundreds Ohms resistor) for 2 clock periods. Then, if the port is externally connected to a low level, it is only kept high with a weak pull up (around $100 \mathrm{~K} \Omega$ ), and if not, the high level is latched high thanks to a medium pull (around $10 \mathrm{k} \Omega$ ).
Thus, when the port is configured as an input, and when this input has been read at a low level, there is a pull-up of around $100 \mathrm{~K} \Omega$, which is quite high, to quickly load the SDA capacitance. So in order to help the reading of a high level just after the reading of a low level, it is possible to force a transition of the SDA port from an input state (1), to an output low state (0), followed by a new transition from this output low state to input state; In this case, the high pull-up has been replaced with a low pull-up which warranties a good reading of the data.

Electrical Characteristics

## Absolute Maximum Ratings ${ }^{(*)}$

| Ambient Temperature Under Bias: |  |
| :---: | :---: |
| $\mathrm{C}=$ commercial............................................. $0^{\circ} \mathrm{C}$ to $70^{\circ} \mathrm{C}$ |  |
| I = industrial .................................................. $-40^{\circ} \mathrm{C}$ to $85^{\circ} \mathrm{C}$ |  |
| Storage Temperature ............................... $65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$ |  |
|  |  |
| Voltage on Any Pin to $\mathrm{V}_{\mathrm{SS}} \ldots . . . . . . . . . . . . . . . . . .-0.5 \mathrm{~V}$ to $\mathrm{V}_{\mathrm{CC}}+0.5 \mathrm{~V}$ Power Dissipation. $\qquad$ 1 W |  |
|  |  |
| Electro-static discharge voltage 1500 V |  |

*NOTICE: Stresses at or above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
Power Dissipation value is based on the maximum allowable die temperature and the thermal resistance of the package

## Power Consumption Measurement

Since the introduction of the first C51 devices, every manufacturer made operating Icc measurements under reset, which made sense for the designs where the CPU was running under reset. In our new devices, the CPU is no longer active during reset, so the power consumption is very low and this is not really representative of what will happen in the customer's system. Thus, while keeping measurements under Reset, we present a new way to measure the operating Icc:

Using an internal test ROM, the following code is executed:
Label: SJMP Label (80 FE)
Ports 3, 4 are disconnected, RST = Vcc, XTAL2 is not connected and XTAL1 is driven by the clock.
This is much more representative of the real operating Icc.
$\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V} ; \mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to $3.6 \mathrm{~V} ; \mathrm{F}=0$ to 24 MHz .
$\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V} ; \mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to $3.6 \mathrm{~V} ; \mathrm{F}=0$ to 24 MHz .

## DC Parameters for Low Voltage

Table 1. DC Parameters for Low Voltage

| Symbol | Parameter | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {IL }}$ | Input Low Voltage | -0.5 |  | 0.8 | V |  |
| $\mathrm{V}_{\mathrm{IH}}$ | Input High Voltage except XTAL1, RST | 2 |  | $\mathrm{V}_{\mathrm{CC}}+0.5$ | V |  |
| $\mathrm{V}_{\mathrm{HH} 1}$ | Input High Voltage, XTAL1, RST | $0.7 \mathrm{~V}_{\mathrm{CC}}$ |  | $\mathrm{V}_{\text {CC }}+0.5$ | V |  |
| $\mathrm{V}_{\text {OL }}$ | Output Low Voltage, ports 3, $4^{(6)}$ |  |  | $\begin{gathered} 0.3 \\ 0.45 \\ 1.0 \end{gathered}$ | $\begin{aligned} & \text { V } \\ & \text { v } \\ & \text { V } \end{aligned}$ | $\begin{aligned} & \mathrm{I}_{\mathrm{OL}}=100 \mu \mathrm{~A} \\ & \mathrm{I}_{\mathrm{OL}}=1.6 \mathrm{~mA} \\ & \mathrm{I}_{\mathrm{OL}}=3.2 \mathrm{~mA} \end{aligned}$ |
| $\mathrm{V}_{\mathrm{OH}}$ | Output High Voltage, ports 3, 4. ${ }^{(6)}$ | $\begin{gathered} 0.9 \mathrm{~V}_{\mathrm{CC}} \\ \mathrm{~V}_{\mathrm{CC}}-0.7 \\ \mathrm{~V}_{\mathrm{CC}}-1.4 \end{gathered}$ |  |  | v v v | $\begin{aligned} & \mathrm{I}_{\mathrm{OH}}=-10 \mu \mathrm{~A} \\ & \mathrm{I}_{\mathrm{OH}}=-30 \mu \mathrm{~A} \\ & \mathrm{I}_{\mathrm{OH}}=-50 \mu \mathrm{~A} \end{aligned}$ |

Table 1. DC Parameters for Low Voltage (Continued)

| Symbol | Parameter | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{V}_{\text {OH2 }}$ | Output High Voltage, ports 3, 4. ${ }^{(6)}$ mode Push pull | $\begin{gathered} 0.9 \mathrm{~V}_{\mathrm{CC}} \\ \mathrm{~V}_{\mathrm{CC}}-0.7 \\ \mathrm{~V}_{\mathrm{Cc}}-1.5 \end{gathered}$ |  |  | $\begin{aligned} & \mathrm{V} \\ & \mathrm{~V} \\ & \mathrm{~V} \end{aligned}$ | $\begin{aligned} & \mathrm{I}_{\mathrm{OH}}=-100 \mu \mathrm{~A} \\ & \mathrm{I}_{\mathrm{OH}}=-1 \mathrm{~mA} \\ & \mathrm{I}_{\mathrm{OH}}=-2 \mathrm{~mA} \end{aligned}$ |
| $1 / 1$ | Logical 0 Input Current ports 3 and $4^{(7)}$ |  |  | -50 | $\mu \mathrm{A}$ | $\mathrm{Vin}=0.45 \mathrm{~V}$ |
| IIL | Input Leakage Current |  |  | $\pm 10$ | $\mu \mathrm{A}$ | $0.45 \mathrm{~V}<\mathrm{Vin}<\mathrm{V}_{\mathrm{CC}}$ |
| $\mathrm{I}_{\text {TL }}$ | Logical 1 to 0 Transition Current, ports 3, $4^{(8)}$ |  |  | -650 | $\mu \mathrm{A}$ | $\mathrm{Vin}=2.0 \mathrm{~V}$ |
| $\mathrm{R}_{\text {RST }}$ | RST Pull up Resistor | 50 | $90^{(5)}$ | 200 | $\mathrm{k} \Omega$ |  |
| ClO | Capacitance of I/O Buffer |  |  | 10 | pF | $\begin{aligned} & \mathrm{Fc}=1 \mathrm{MHz} \\ & \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C} \end{aligned}$ |
| $\mathrm{I}_{\text {PD }}$ | Power Down Current |  | 50 | 200 | $\mu \mathrm{A}$ | $\mathrm{V}_{\mathrm{CC}}=3.0 \mathrm{~V}$ to $3.6 \mathrm{~V}^{(3)}$ |
| $\mathrm{F}_{\text {OSCB }}$ | OSCB unlocked frequency | 10.8 | 12 | 13.2 | MHz | With ideal R and C |
| $\mathrm{F}_{\text {OSCB }}$ | OSCB locked frequency | 11.5 | 12 | 12.5 | MHz | With ideal R and C |
| $\mathrm{F}_{\text {Osc }}$ | OSCC frequency | 8.4 | 14 | 19.6 | MHz |  |
| $\mathrm{I}_{\mathrm{Cc}}$ under RESET | Power Supply Current Maximum values, X1 mode, OSCA oscillator ${ }^{(9)}$ |  |  | 4 | mA | $\begin{aligned} & \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}^{(1)} \\ & \text { OSCA + Prescaler } \end{aligned}$ |
| $l_{c c}$ operating | Power Supply Current Maximum values, X1 mode, OSCA oscillator ${ }^{(9)}$ |  |  | $0.4 *$ + ${ }^{\text {a }}$ | mA | $\begin{aligned} & V_{\mathrm{CC}}=3.3 V^{(10)} \\ & \mathrm{F} \text { in } \mathrm{MHz} \end{aligned}$ |
| $\begin{aligned} & \mathrm{I}_{\mathrm{CC}} \\ & \text { idle } \end{aligned}$ | Power Supply Current Maximum values, X1 mode, OSCA oscillator ${ }^{(9)}$ |  |  | 6 | mA | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}^{(2)}$ |
| $I_{C c}$ under RESET | Power Supply Current Maximum values, X1 mode, OSCB oscillator ${ }^{(9)}$ |  |  | 900 | uA | $\begin{aligned} & \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}^{(1)} \\ & \mathrm{OSCB}+\text { Prescaler } \end{aligned}$ |
| $I_{c c}$ operating | Power Supply Current Maximum values, X1 mode, OSCB oscillator ${ }^{(9)}$ |  |  | 5 | mA | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}^{(10)}$ |
| $\begin{aligned} & \mathrm{I}_{\mathrm{CC}} \\ & \text { idle } \end{aligned}$ | Power Supply Current Maximum values, X1 mode, OSCB oscillator ${ }^{(9)}$ |  |  | 4.8 | mA | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}^{(2)}$ |
| $I_{C C}$ under RESET | Power Supply Current Maximum values, X1 mode, OSCC oscillator ${ }^{(9)}$ |  |  | 650 | $\mu \mathrm{A}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}^{(1)} \\ & \mathrm{OSCC}+\text { Prescaler } \end{aligned}$ |
| $I_{c c}$ operating | Power Supply Current Maximum values, X1 mode, OSCC oscillator ${ }^{(9)}$ |  |  | 5 | mA | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}^{(10)}$ |
| $\begin{aligned} & \mathrm{I}_{\mathrm{CC}} \\ & \text { idle } \end{aligned}$ | Power Supply Current Maximum values, X1 mode, OSCC oscillator ${ }^{(9)}$ |  |  | 4.8 | mA | $\mathrm{V}_{\mathrm{CC}}=3.3 \mathrm{~V}^{(2)}$ |
| $V_{\text {RET }}$ | Supply voltage during power down mode | 2.7 |  |  | V |  |
| VPFDP | Power fail high level threshold | 2.6 | 2.8 | 2.95 | V |  |
| VPFDM | Power fail low level threshold (default) | 2.45 | 2.55 | 2.7 | V |  |
|  | Power fail hysteresis VPFDP - VPFDM | 150 | 250 | 350 | mV |  |
| $t_{G}$ | Glitch maximum time |  | 100 |  | ns | Vcc down to 2.5 V |

Table 1. DC Parameters for Low Voltage (Continued)

| Symbol | Parameter | Min | Typ | Max | Unit | Test Conditions |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $t_{R}$ | Supply rise time | $1 u s$ |  | 1 s |  |  |

Notes: 1. $\mathrm{I}_{\mathrm{CC}}$ under reset is measured with all output pins disconnected; XTAL1 driven with $\mathrm{T}_{\mathrm{CLCH}}, \mathrm{T}_{\mathrm{CHCL}}=5 \mathrm{~ns}$ (see Figure 42.), $\mathrm{V}_{\mathrm{IL}}=$ $\mathrm{V}_{\mathrm{SS}}+0.5 \mathrm{~V}$,
$\mathrm{V}_{\mathrm{IH}}=\mathrm{V}_{\mathrm{CC}}-0.5 \mathrm{~V}$; XTAL2 N.C.; $\mathrm{Vpp}=\mathrm{RST}=\mathrm{V}_{\mathrm{CC}}$. $\mathrm{I}_{\mathrm{CC}}$ would be slightly higher if a crystal oscillator used
2. Idle $\mathrm{I}_{\mathrm{CC}}$ is measured with all output pins disconnected; XTAL1 driven with $\mathrm{T}_{\mathrm{CLCH}}, \mathrm{T}_{\mathrm{CHCL}}=5 \mathrm{~ns}, \mathrm{~V}_{\mathrm{IL}}=\mathrm{V}_{\mathrm{SS}}+0.5 \mathrm{~V}, \mathrm{~V}_{\mathrm{IH}}=\mathrm{V}_{\mathrm{CC}}-$ 0.5 V ; XTAL2 N.C; $\mathrm{Vpp}=\mathrm{RST}=\mathrm{V}_{\mathrm{SS}}$ (see Figure 40.).
3. Power Down $I_{C C}$ is measured with all output pins disconnected; $\mathrm{Vpp}=\mathrm{V}_{\mathrm{SS}} ; \mathrm{XTAL2} \mathrm{NC}$.; $R S T=\mathrm{V}_{\mathrm{cc}}$ (see Figure 41.).
4. Not Applicable
5. Typical are based on a limited number of samples and are not guaranteed. The values listed are at room temperature and 3.3V.
6. If $\mathrm{I}_{\mathrm{OL}}$ exceeds the test condition, $\mathrm{V}_{\mathrm{OL}}$ may exceed the related specification. Pins are not guaranteed to sink current greater than the listed test conditions.
7. For other values, please contact your sales office.
8. When port configuration have weak pull-up activated.
9. When port configuration is quasi-bidirectional.
10. Operating $\mathrm{I}_{\mathrm{CC}}$ is measured with all output pins disconnected; XTAL1 driven with $\mathrm{T}_{\mathrm{CLCH}}, \mathrm{T}_{\mathrm{CHCL}}=5 \mathrm{~ns}$ (see Figure 42.), $\mathrm{V}_{\mathrm{IL}}=$ $\mathrm{V}_{\mathrm{SS}}+0.5 \mathrm{~V}$,
$\mathrm{V}_{\mathrm{IH}}=\mathrm{V}_{\mathrm{CC}}-0.5 \mathrm{~V}$; XTAL2 N.C.; $\overline{\mathrm{RST}}=\mathrm{V}_{\mathrm{CC}}$; The internal ROM runs the code 80 FE (label: SJMP label). $\mathrm{I}_{\mathrm{CC}}$ would be slightly higher if a crystal oscillator is used. Measurements are made with OTP products when possible, which is the worst case.

Figure 38. $\mathrm{I}_{\mathrm{CC}}$ Test Condition, under reset


All other pins are disconnected.

Figure 39. Operating $\mathrm{I}_{\mathrm{CC}}$ Test Condition


All other pins are disconnected.

Figure 40. $\mathrm{I}_{\mathrm{CC}}$ Test Condition, Idle Mode


All other pins are disconnected.

Figure 41. $\mathrm{I}_{\mathrm{CC}}$ Test Condition, Power-Down Mode


All other pins are disconnected.

Figure 42. Clock Signal Waveform for $\mathrm{I}_{\mathrm{CC}}$ Tests in Active and Idle Modes

$$
\begin{gathered}
0.7 \mathrm{~V}_{\mathrm{CC}} \\
0.2 \mathrm{~V}_{\mathrm{CC}}-0.1 \\
\mathrm{~V}_{\mathrm{CC}}-0.5 \mathrm{~V} \\
\mathrm{~T}_{\mathrm{CHCL}} \\
\mathrm{~T}_{\mathrm{CLCH}}=\mathrm{T}_{\mathrm{CHCL}}=5 \mathrm{~ns} .
\end{gathered}
$$

## DC Parameters for A/D Converter

$\mathrm{T}_{\mathrm{A}}=0^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V} ; \mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to $3,6 \mathrm{~V} ; \mathrm{F}=0$ to 24 MHz .
$\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C} ; \mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V} ; \mathrm{V}_{\mathrm{CC}}=3 \mathrm{~V}$ to $3,6 \mathrm{~V} ; \mathrm{F}=0$ to 24 MHz .
Table 2. DC Parameters for Low Voltage

| Symbol | Parameter | Min | Typ | Max | Unit | Test Conditions |
| :---: | :--- | :---: | :---: | :---: | :---: | :--- |
|  | Resolution |  | 10 |  | bit |  |
| AVin | Analog input voltage | Vss -0.2 |  | Vcc +0.2 | V |  |
| Rref | Resistance between Vref and Vss | 13 | 18 | 24 | KO <br> hm |  |
| Vref | Value of integrated voltage source | 2.30 | 2.40 | 2.50 | V |  |
| Vref <br> drift | Vref Voltage drift over temperature |  |  | 150 | $\mathrm{uV} /$ <br> ${ }^{\circ} \mathrm{C}$ |  |
| Lref | Load on integrated voltage source | 10 |  |  | KO <br> hm |  |
| Cai | Analog input Capacitance |  | 60 |  | pF | During sampling |
|  | Integral non linearity |  | 1 | 2 | Isb | With ideal <br> external Ref (1) |
|  | Differential non linearity | 0.5 | 1 | Isb |  |  |
|  | Offset error | -2 |  | 2 | Isb |  |
|  | Input source impedance |  | 1 | KO <br> hm | For 10 bit <br> resolution at <br> maximum speed |  |

Note: (1) With $1 \mathrm{sb}=2.4 / 1024=2.4 \mathrm{mV}$, typical integral linearity is:

$$
\frac{(2,4-V r e f)}{2,4 \times 10^{-3}}
$$

## AC Parameters

## Explanation of the AC Symbols

Each timing symbol has 5 characters. The first character is always a " T " (stands for Time). The other characters, depending on their positions, stand for the name of a signal or the logical status of that signal. The following is a list of all the characters and what they stand for.

Example: $\mathrm{T}_{\mathrm{XHDV}}=$ Time from clock rising edge to input data valid.
$\mathrm{T}_{\mathrm{A}}=0$ to $+70^{\circ} \mathrm{C}$ (commercial temperature range); $\mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V} ; 3 \mathrm{~V}<\mathrm{V}_{\mathrm{CC}}<3.6 \mathrm{~V} ;-\mathrm{L}$ range. $\mathrm{T}_{\mathrm{A}}=-40^{\circ} \mathrm{C}$ to $+85^{\circ} \mathrm{C}$ (industrial temperature range); $\mathrm{V}_{\mathrm{SS}}=0 \mathrm{~V} ; 3 \mathrm{~V}<\mathrm{V}_{\mathrm{CC}}<3.6 \mathrm{~V} ;-\mathrm{L}$ range.

Table 3. gives the maximum applicable load capacitance for Port 1, 3 and 4. Timings will be guaranteed if these capacitances are respected. Higher capacitance values can be used, but timings will then be degraded.
Table 3. Load Capacitance versus speed range, in pF

|  | - L |
| :---: | :---: |
| Port 3 \& 4 | 60 |

Table 4. Max frequency for Derating Formula Regarding the Speed Grade

|  | -L X1 mode | -L X2 mode |
| :---: | :---: | :---: |
| Freq (MHz) | $40^{(1)}$ | 20 |
| $\mathbf{T}(\mathbf{n s})$ | 25 | 50 |

1. Oscillator speed is limited to 24 Mhz

## External Clock Drive

Characteristics (XTAL1)

| Symbol | Parameter | Min | Max | Units |
| :---: | :--- | :---: | :---: | :---: |
| $\mathrm{T}_{\mathrm{CLCL}}$ | Oscillator Period | 25 |  | ns |
| $\mathrm{~T}_{\mathrm{CHCX}}$ | High Time | 5 |  | ns |
| $\mathrm{~T}_{\mathrm{CLCX}}$ | Low Time | 5 | 5 | ns |
| $\mathrm{~T}_{\mathrm{CLCH}}$ | Rise Time |  | 5 | ns |
| $\mathrm{~T}_{\mathrm{CHCL}}$ | Fall Time | 40 | 60 | ns |
| $\mathrm{~T}_{\mathrm{CHCX}} / \mathrm{T}_{\mathrm{CLCX}}$ | Cyclic ratio in X2 mode |  | $\%$ |  |

Figure 43. External Clock Drive Waveforms


## A/D Converter

| Symbol | Parameter | Min | Typ | Max | Units |
| :--- | :--- | :---: | :---: | :---: | :---: |
|  | Conversion time |  | 11 |  | Clock periods (1 for <br> sampling, 10 for <br> conversion) |
| FConv_Ck | Clock Conversion frequency |  |  | $550(1)$ | kHz |
|  | Sampling frequency | 10 |  | 50 | kilo samples per <br> second |

Notes: 1. For 10 bits resolution

PWM Outputs

| Symbol | Parameter | Min | Typ | Max | Units |
| :---: | :--- | :---: | :---: | :---: | :---: |
| $\operatorname{Tr}$ | Rise time of PWM outputs |  | 60 |  | ns (load 300 pF ) <br> Can be slower |
| Tf | Fall time of PWM outputs |  | 30 |  | ns (300 pF) <br> Can be slower |
|  |  |  |  |  |  |
|  |  |  |  |  |  |

AC Testing Input/Output Waveforms

Figure 44. AC Testing Input/Output Waveforms


Float Waveforms

AC inputs during testing are driven at $\mathrm{V}_{\mathrm{CC}}-0.5$ for a logic " 1 " and 0.45 V for a logic " 0 ". Timing measurement are made at $\mathrm{V}_{\mathrm{IH}}$ min for a logic " 1 " and $\mathrm{V}_{\mathrm{IL}}$ max for a logic " 0 ".

Figure 45. Float Waveforms


For timing purposes as port pin is no longer floating when a 100 mV change from load voltage occurs and begins to float when a 100 mV change from the loaded $\mathrm{V}_{\mathrm{OH}} / \mathrm{V}_{\mathrm{OL}}$ level occurs. $\mathrm{I}_{\mathrm{OL}} / \mathrm{I}_{\mathrm{OH}} \geq \pm 20 \mathrm{~mA}$.

Valid in normal clock mode. In X2 mode XTAL2 signal must be changed to XTAL2 divided by two.

Figure 46. Clock Waveforms


PORT OPERATION
$\xrightarrow{\text { OLD DATA } \sqrt{n}}$


This diagram indicates when signals are clocked internally. The time it takes the signals to propagate to the pins, however, ranges from 25 to 125 ns . This propagation delay is dependent on variables such as temperature and pin loading. Propagation also varies from output to output and component. Typically though ( $T_{A}=25^{\circ} \mathrm{C}$ fully loaded) $\overline{\mathrm{RD}}$ and $\overline{W R}$ propagation delays are approximately 50 ns . The other signals are typically 85 ns . Propagation delays are incorporated in the AC specifications.

## Typical Application

Figure 47. Typical Application Diagram


Ordering Information
Table 7. Possible Order Entries

| Part Number | Memory <br> Size | Supply <br> Voltage | Temperature <br> Range | Max Frequency | Package | Packing |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| AT83EB5114xxxTGRIL | 4 Kb ROM | 3 to 3.6 V | Industrial | 40 MHz | S020 | Reel |
| AT89EB5114-TGSIL | 4 Kb Flash | 3 to 3.6 V | Industrial | 40 MHz | SO20 | Stick |

## Package Drawings

SO20


|  | MM |  | I NCH |  |
| :---: | :---: | :---: | :---: | :---: |
| A | 2. 35 | 2. 65 | . 093 | 104 |
| A1 | 0.10 | 0.30 | . 004 | 012 |
| B | 0.35 | 0.49 | . 014 | 019 |
| C | 0. 23 | 0.32 | . 009 | 013 |
| D | 12.60 | 13.00 | . 496 | 512 |
| E | 7. 40 | 7. 60 | . 291 | 299 |
| e | 1.27 | BSC | . 050 | BSC |
| H | 10.00 | 10.65 | . 394 | 419 |
| h | 0.25 | 0.75 | . 010 | 029 |
| L | 0.40 | 1.27 | . 016 | 050 |
| N | 20 |  | 20 |  |
| a | $0^{\circ}$ |  | $8^{\circ}$ |  |

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